

MD918, A, B (SILICON)

MD918F, AF, BF

MULTIPLE SILICON ANNULAR TRANSISTORS

... designed for use as differential amplifiers, dual high frequency amplifiers, front end detectors and temperature compensation applications.

- Low Collector-Emitter Saturation Voltage – $V_{CE(sat)} = 0.2 \text{ Vdc (Max) @ } I_C = 10 \text{ mAdc}$
- DC Current Gain – 50 (Min) @ $I_C = 3.0 \text{ mAdc}$
- High Current-Gain – Bandwidth Product – $f_T = 600 \text{ MHz @ } I_C = 4.0 \text{ mAdc}$

MAXIMUM RATINGS

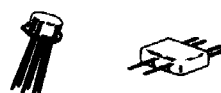
Rating	Symbol	Value	Unit	
Collector-Emitter Voltage	V_{CEO}	15	Vdc	
Collector-Base Voltage	V_{CB}	30	Vdc	
Emitter-Base Voltage	V_{EB}	3.0	Vdc	
Collector Current – Continuous	I_C	50	mAdc	
		One Die	All Die	
Total Power Dissipation @ $T_A = 25^\circ\text{C}$ MD918,A,B MD918F,AF,BF	P_D	550 350	600 400	mW
Derate Above 25°C				
MD918,A,B MD918F,AF,BF		3.14 2.0	3.42 2.28	mW/ $^\circ\text{C}$
Total Power Dissipation @ $T_C = 25^\circ\text{C}$ MD918,A,B MD918F,AF,BF	P_D	1.4 0.7	2.0 1.4	Watts
Derate Above 25°C				
MD918,A,B MD918F,AF,BF		8.0 4.0	11.4 8.0	mW/ $^\circ\text{C}$
Operating and Storage Junction Temperature Range	T_J, T_{stg}	-65 to +200		$^\circ\text{C}$

THERMAL CHARACTERISTICS

Characteristic	Symbol	One Die	All Die Equal Power	Unit
Thermal Resistance, Junction to Ambient MD918,A,B MD918F,AF,BF	$R_{\theta JA(1)}$	319 500	292 438	$^\circ\text{C/W}$
Thermal Resistance, Junction to Case MD918,A,B MD918F,AF,BF	$R_{\theta JC}$	125 260	87.5 125	$^\circ\text{C/W}$
		Junction to Ambient	Junction to Case	
Coupling Factors MD918,A,B MD918F,AF,BF		83 75	40 0	%

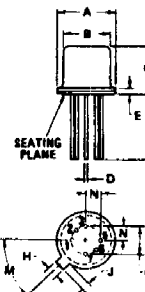
(1) $R_{\theta JA}$ is measured with the device soldered into a typical printed circuit board.

NPN SILICON MULTIPLE TRANSISTORS



MD918
MD918A
MD918B

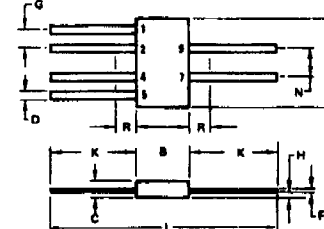
STYLE 1:
PIN 1: COLLECTOR
2: BASE
3: EMITTER
4: OMITTED
5: EMITTER
6: BASE
7: COLLECTOR
8: OMITTED



DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	8.51	9.40	0.335	0.370
B	7.75	8.51	0.305	0.335
C	2.81	3.70	0.110	0.146
D	0.41	0.53	0.016	0.021
E	0.30 BSC			
F	0.20 BSC			
G	0.71	0.86	0.028	0.034
H	0.74	1.14	0.029	0.045
J	12.70	-	0.500	-
K	40 BSC			
N	2.54 BSC			

CASE 054-07

MD918F
MD918AF
MD918BF



DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	6.10	7.36	0.240	0.290
B	2.82	4.06	0.110	0.160
C	0.78	2.03	0.030	0.080
D	0.28	0.48	0.011	0.019
E	0.10 BSC			
F	0.10 BSC			
G	1.27 BSC	-	0.050 BSC	-
H	-	0.89	-	0.035
J	0.51	-	0.020	-
K	2.54 BSC			
L	-	1.27	-	0.050

STYLE 1:
PIN 1: BASE
2: EMITTER
4: EMITTER
5: BASE
7: COLLECTOR
8: COLLECTOR

CASE 610A-03



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MD918,A,B, MD918F,AF,BF (continued)

THERMAL COUPLING AND EFFECTIVE THERMAL RESISTANCE

In multiple chip devices, coupling of heat between die occurs. The junction temperature can be calculated as follows:

(1) $\Delta T_{J1} = R_{\theta 1} P_{D1} + R_{\theta 2} K_{\theta 2} P_{D2}$

Where ΔT_{J1} is the change in junction temperature of die 1
 $R_{\theta 1}$ and $R_{\theta 2}$ is the thermal resistance of die 1 and die 2
 P_{D1} and P_{D2} is the power dissipated in die 1 and die 2
 $K_{\theta 2}$ is the thermal coupling between die 1 and die 2

An effective package thermal resistance can be defined as follows:

(2) $R_{\theta(EFF)} = \Delta T_{J1} / P_{DT}$

where: P_{DT} is the total package power dissipation.
 Assuming equal thermal resistance for each die, equation (1) simplifies to

(3) $\Delta T_{J1} = R_{\theta 1} (P_{D1} + K_{\theta 2} P_{D2})$

For the conditions where $P_{D1} = P_{D2}$, $P_{DT} = 2P_D$, equation (3) can be further simplified and by substituting into equation (2) results in

(4) $R_{\theta(EFF)} = R_{\theta 1} (1 + K_{\theta 2}) / 2$

Values for the coupling factors when either the case or the ambient is used as a reference are given in the table on page 1.

ELECTRICAL CHARACTERISTICS ($T_A = 25^\circ\text{C}$ unless otherwise noted.)

Characteristic	Symbol	Min	Typ	Max	Unit
OFF CHARACTERISTICS					
Collector-Emitter Breakdown Voltage ⁽¹⁾ ($I_C = 3.0 \text{ mAdc}$, $I_B = 0$)	BV_{CEO}	15	—	—	Vdc
Collector-Base Breakdown Voltage ($I_C = 1.0 \mu\text{Adc}$, $I_E = 0$)	BV_{CBO}	30	—	—	Vdc
Emitter-Base Breakdown Voltage ($I_E = 10 \mu\text{Adc}$, $I_C = 0$)	BV_{EBO}	3.0	—	—	Vdc
Collector Cutoff Current ($V_{CB} = 15 \text{ Vdc}$, $I_E = 0$) ($V_{CB} = 15 \text{ Vdc}$, $I_E = 0$, $T_A = 150^\circ\text{C}$)	I_{CBO}	—	—	10 1.0	nAdc μAdc
ON CHARACTERISTICS					
DC Current Gain ($I_C = 3.0 \text{ mAdc}$, $V_{CE} = 5.0 \text{ Vdc}$)	h_{FE}	50	165	—	—
Collector-Emitter Saturation Voltage ($I_C = 10 \text{ mAdc}$, $I_B = 1.0 \text{ Adc}$)	$V_{CE(sat)}$	—	0.09	0.2	Vdc
Base-Emitter Saturation Voltage ($I_C = 10 \text{ mAdc}$, $I_B = 1.0 \text{ mAdc}$)	$V_{BE(sat)}$	—	0.86	0.9	Vdc
DYNAMIC CHARACTERISTICS					
Current-Gain - Bandwidth Product ($I_C = 4.0 \text{ mAdc}$, $V_{CE} = 10 \text{ Vdc}$, $f = 100 \text{ MHz}$) ^a	f_T	600	1150	—	MHz
Output Capacitance ($V_{CB} = 10 \text{ Vdc}$, $I_E = 0$, $f = 100 \text{ kHz}$)	C_{ob}	—	1.1	1.7	pF
Input Capacitance ($V_{BE} = 0.5 \text{ Vdc}$, $I_C = 0$, $f = 100 \text{ kHz}$)	C_{ib}	—	1.15	2.0	pF
Noise Figure ($I_C = 1.0 \text{ mAdc}$, $V_{CE} = 6.0 \text{ Vdc}$, $R_S = 400 \Omega$, $f = 60 \text{ MHz}$)	NF	—	—	6.0	dB
MATCHING CHARACTERISTICS					
DC Current-Gain Ratio ⁽²⁾ ($I_C = 1.0 \text{ mAdc}$, $V_{CE} = 5.0 \text{ Vdc}$)	h_{FE1}/h_{FE2}	0.8 0.9	—	1.0 1.0	—
Base-Emitter Voltage Differential ($I_C = 1.0 \text{ mAdc}$, $V_{CE} = 5.0 \text{ Vdc}$)	$ V_{BE1} - V_{BE2} $	—	—	10 5.0	mVdc
Base-Emitter Voltage Differential Gradient ($I_C = 1.0 \text{ mAdc}$, $V_{CE} = 5.0 \text{ Vdc}$, $T_A = -55 \text{ to } +125^\circ\text{C}$)	$\frac{\Delta V_{BE1} - V_{BE2} }{\Delta T_A}$	—	—	20 10	$\mu\text{V/dc}$ $^\circ\text{C}$

(1) Pulse Test: Pulse Width $\leq 300 \mu\text{s}$, Duty Cycle $\leq 2.0\%$.
 (2) The lowest h_{FE} reading is taken as h_{FE1} for this ratio.