

2-GHz Single Balanced Mixer

Description

The U2796B-FP is a 2-GHz down conversion mixer for telecommunication systems, e.g. cellular radio, CT1, CT2, DECT, PCN, using TELEFUNKEN advanced bipolar technology. The U2796B is well suited for the receiver

portion of the RF circuit. Single balanced structure has been chosen for the best noise performance and low current consumption. The IIP3 is programmable.

Features

- Supply voltage range: 2.7 to 5.5 V
- Excellent isolation characteristics
- Low current consumption: 3.2 mA without R_{IP3}
- IIP3 programmable
- Input frequency operating range up to 2 GHz
- RF characteristic nearly independent of supply voltage

Benefits

- Stand alone product
- Low current consumption extends talk time
- 3-V operation requires small space for batteries

Block Diagram

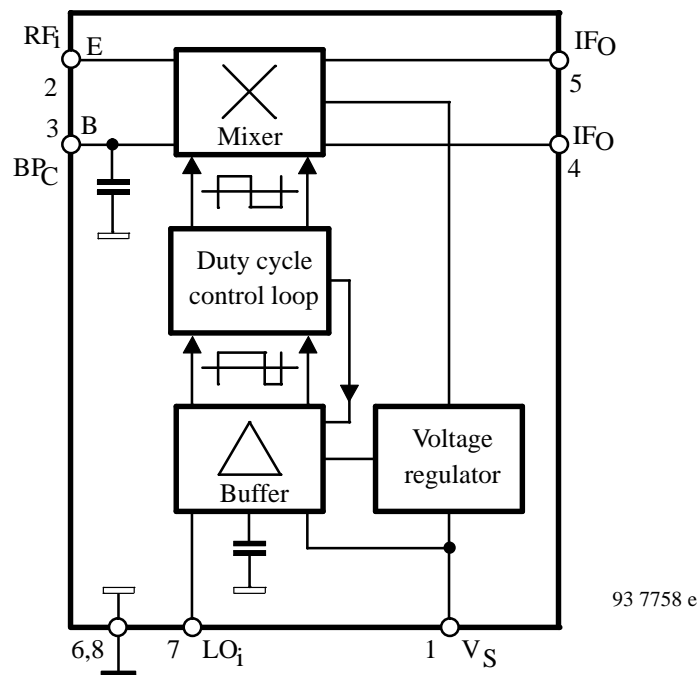
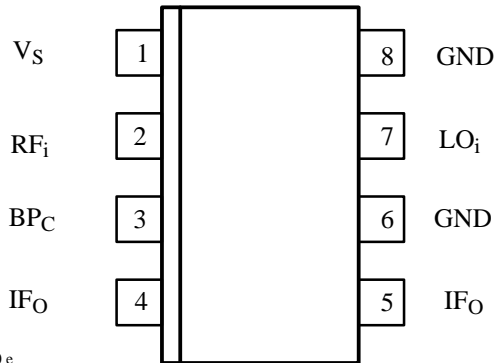


Figure 1.

Pin Description



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Figure 2.

Pin	Symbol	Function
1	V_S	Supply voltage
2	RF	RF input and IIP3 programming port
3	BP_C	By-pass capacitor
4	IF_O	IF output
5	IF_O	IF output
6	GND	Ground
7	LO_i	Local oscillator input
8	GND	Ground

Absolute Maximum Ratings

Parameters	Symbol	Value	Unit	
Supply voltage	Pin 1	V_S	6	V
Input voltage	Pins 2, 3, 4, 5 and 7	V_i	0 to V_S	V
Junction temperature	T_j	125	°C	
Storage temperature	T_{stg}	-40 to +125	°C	

Operating Range

Parameters	Symbol	Value	Unit	
Supply voltage range	Pin 1	V_S	2.7 to 5.5	V
Ambient temperature	T_{amb}	-40 to +85	°C	

Thermal Resistance

Parameters	Symbol	Value	Unit	
Junction ambient	SO 8	R_{thJA}	175	K/W

Electrical Characteristics

Test conditions (unless otherwise specified):

$V_S = 3 \text{ V}$, $f_{LO} = 900 \text{ MHz}$; $I_M = 1.2 \text{ mA}$, $T_{amb} = 25^\circ\text{C}$. System impedance $Z_O = 50 \Omega$

Parameters	Test conditions / Pin	Symbol	Min.	Typ.	Max.	Unit
Supply voltage	Pin 1	V_S	2.7		5.5	V
Supply current	$R_{IP3} = \infty$, Pin 1	I_S	2.8	3.2	3.7	mA
Conversion power gain	$RL = 3 \text{ k}\Omega$, $R_{IP3} = \infty$ $f_{LO} = 900 \text{ MHz}$	PG_C		9		dB
Figure 4	$f_{LO} = 1700 \text{ MHz}$ $f_{IF} = 45 \text{ MHz}$			9		
Isolation						
LO-spurious at RF_{in}	$P_{iLO} = -10 \text{ dBm}$ Figure 5 Pin 7 to 2	IS_{LORF}			-35	dBm
RF to LO	$P_{iRF} = -25 \text{ dBm}$ Pin 2 to 7 $f_{LO} = 900 \text{ MHz}$	IS_{RFLO}	30	40		dB
Figure 6	$f_{LO} = 1700 \text{ MHz}$			20		
Operating frequencies						
RF frequency	Pin 2	RF_i	2000			MHz
LO_{in} frequency	Pin 7	LO_i	2000			MHz
IF_{out} frequency	Pins 4 and 5	IF_o	300			MHz
Input level						
RF input (-1 dB comp.)	$RL = 50 \Omega$, Pin 2	P_{iRF}		-15		dBm
3rd order intercept point	$P_{iLO} = -10 \text{ dBm}$, $R_{IP3} = \infty$ Figure 2 Pin 2	$IIP3$		-4		dBm
LO input	Pin 7	P_{iLO}		-6	0	dBm
Impedances						
RF input	Pin 2	Z_{iRF}		25		Ω
LO input	Pin 7	Z_{iLO}		50		Ω
IF output	Pins 4 and 5	Z_{oIF}		> 10 k Ω // 0.9 pF		
Noise figure (DSB)	$P_{iLO} = 0 \text{ dBm}$, $RL > 3 \text{ k}\Omega$ $f_{LO} = 900 \text{ MHz}$	NF_{50}		9		dB
Figure 7	$f_{LO} = 1700 \text{ MHz}$			12		
Voltage standing wave ratio LO	Pin 7	VSWR- LO		1.3	2	

Note: I_M = Internal mixer current (see figure 2)

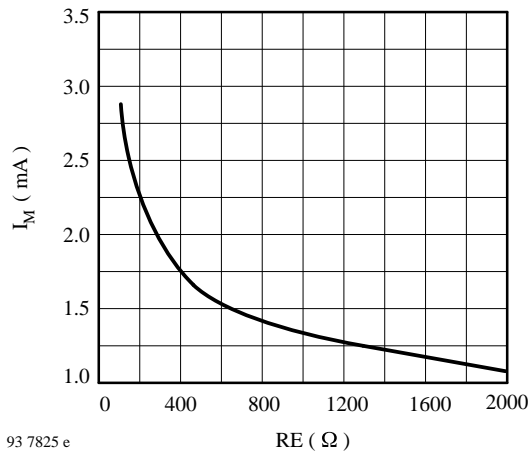


Figure 3. Mixer current (I_M) versus RE

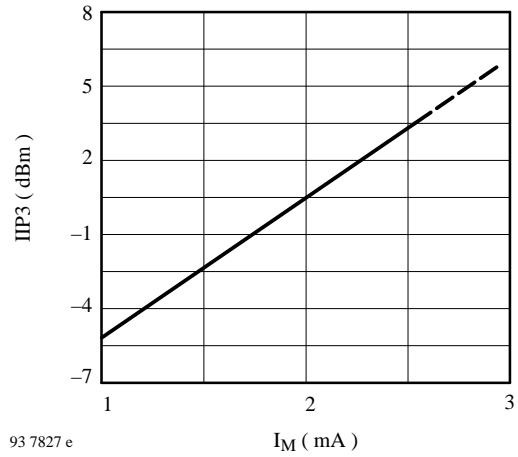


Figure 4. Third-order input intercept IIP3 point versus I_M

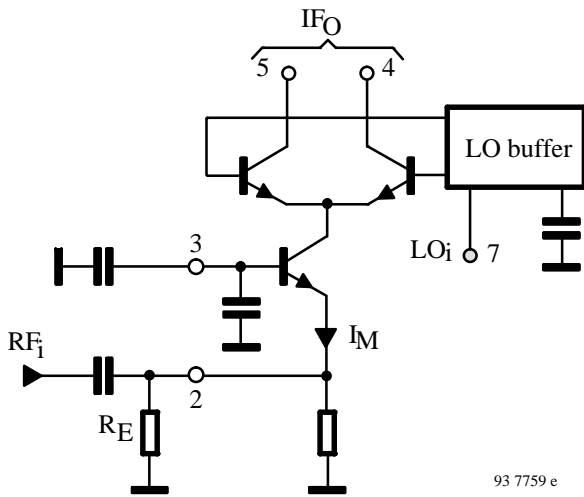


Figure 5. Mixer circuitry

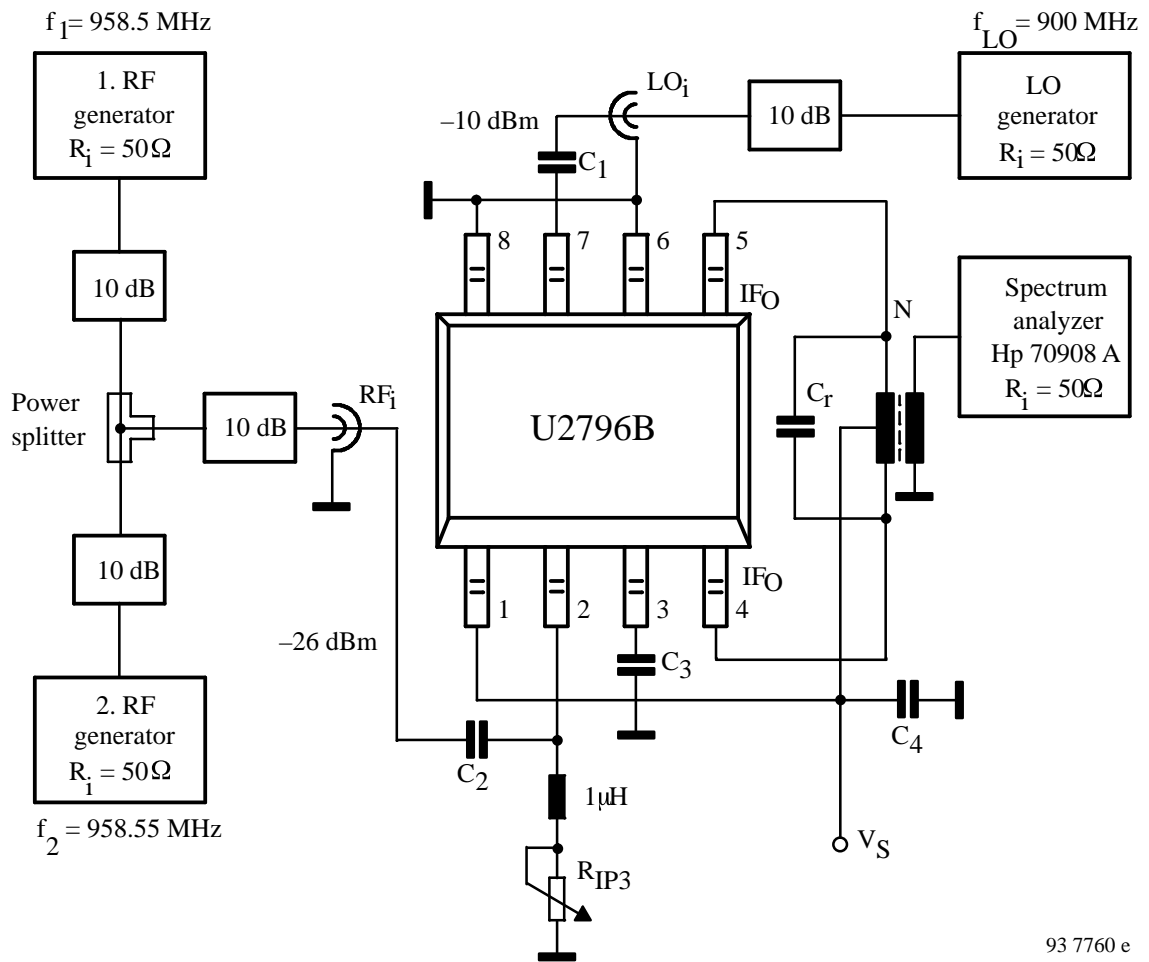


Figure 6. Test circuit-conversion power gain (PG_C) and 3rd order input intercept point (IIP3)

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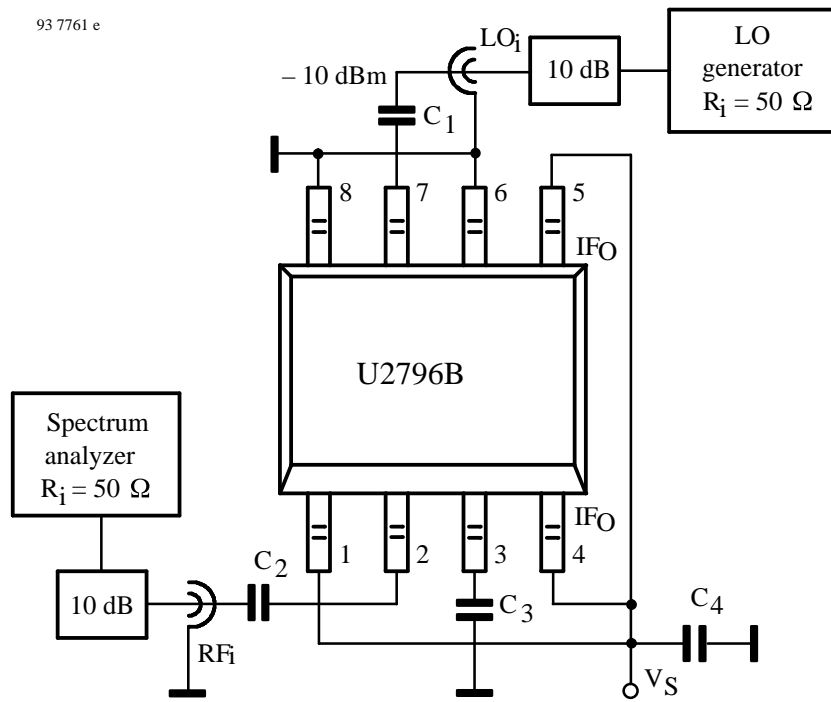
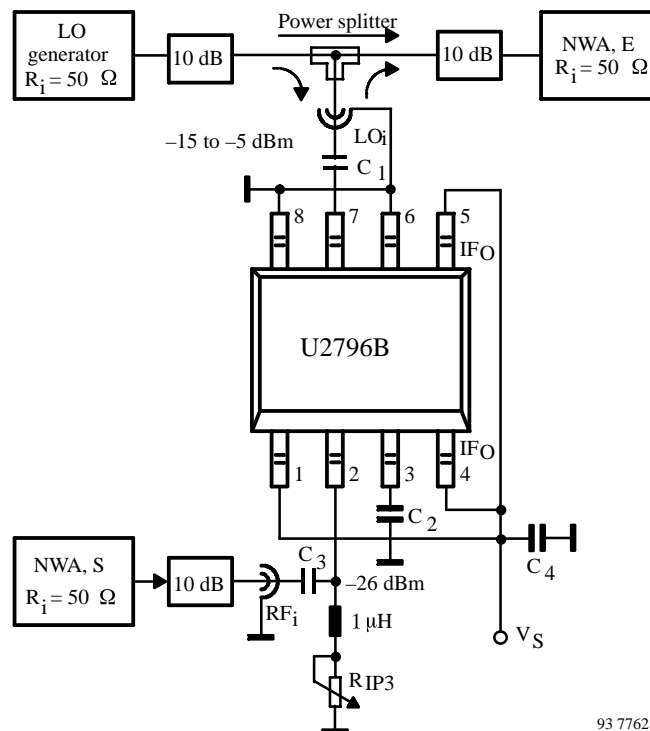
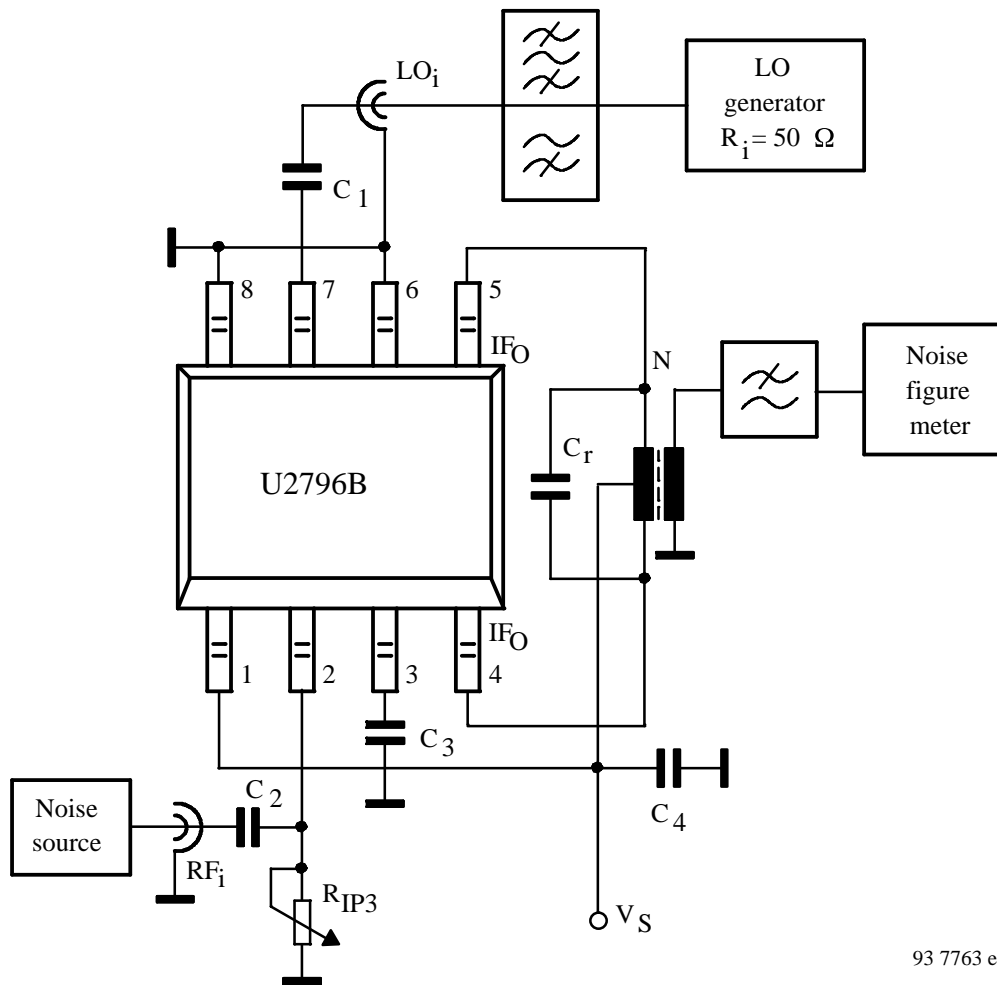


Figure 7. Test circuit-isolation LO to RF



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Figure 8. Test circuit-isolation RF to LO



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Figure 9. Test circuit-noise figure

Note:

1. The noise floor of the LO generator might influence the noise figure test result. In order to avoid this, either a band pass or a high pass filter with $f_c > f_{IF}$ should be implemented.
2. If IF output network does not provide sufficient suppression of the LO component, a low pass filter should be inserted to avoid overdriving the noise figure meter.
3. For best noise performance 0 dBm LO power level is required.

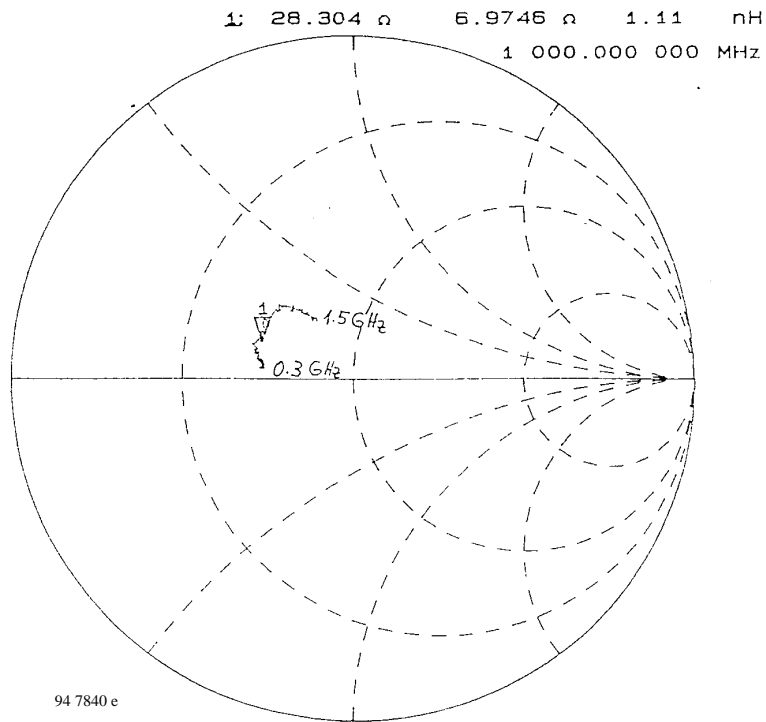


Figure 10. S11 RF input impedance

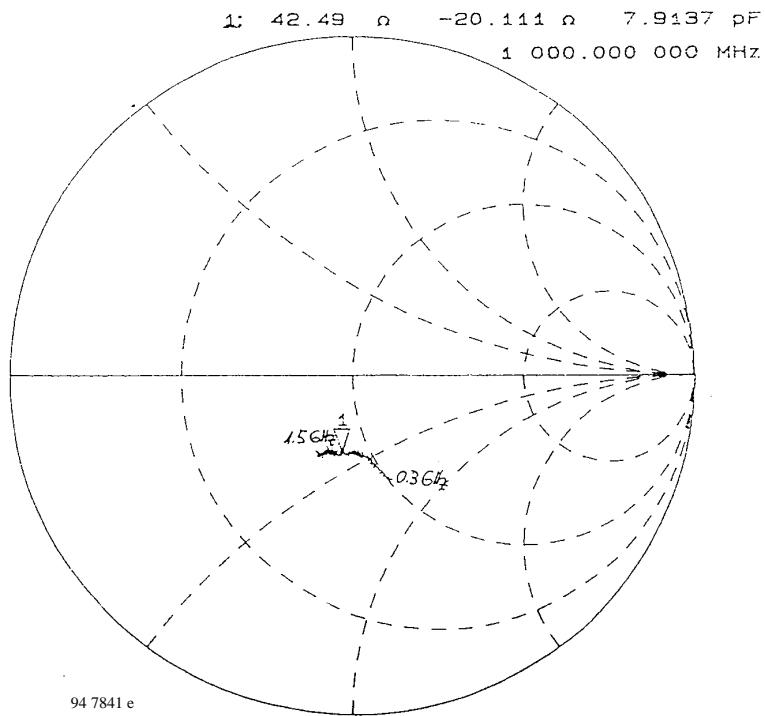


Figure 11. S11 LO input impedance

Application Circuit

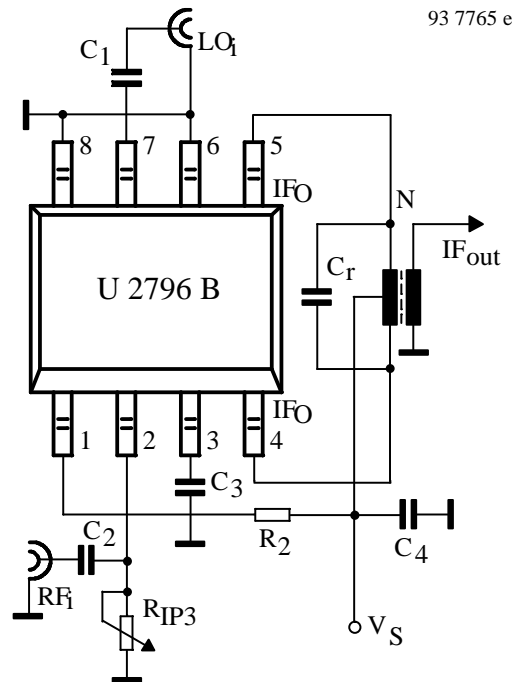


Figure 12.

Recommended Values for the Evaluator

C_1 and $C_2 = 150$ pF, C_3 and $C_4 = 100$ nF. C_r is calculated for resonance with the balun at f_{IF} , or as a high pass filter for f_{LO} . The output balun transformer ratio $> 8:1$ for $Z_O = 50 \Omega$. R_2 increases the IF output level and is calculated from:

$$R_2 = \frac{V_S(4,5) - V_S(1)}{I_S(1)}$$

For example $V_S(4,5) = 4$ V, $V_S(1) = 3$ V, $I_S(1) = 2.2$ mA
 $R_2 \approx 470 \Omega$, where $I_S(1)$ is the current consumption without the mixer stage.

Application Hint

The output transformer at the pins 4 and 5 can be replaced by LC-circuits like one of the following proposals, which are saving space compared to the transformer and are suitable for higher IF frequencies. When applying one of these solutions, it has to be checked whether the requirements on noise figure and gain can be achieved.

The second circuit was dimensioned for approximately 130 MHz and a load resistance of 50Ω . If for instance the impedance of a subsequent filter is 1 k Ω , the capacitive voltage divider may be left out.

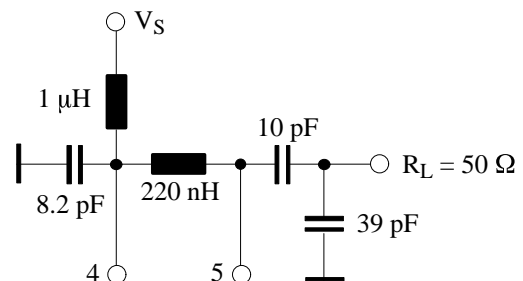
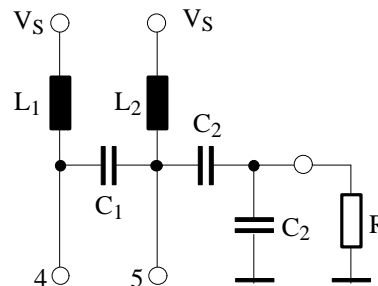
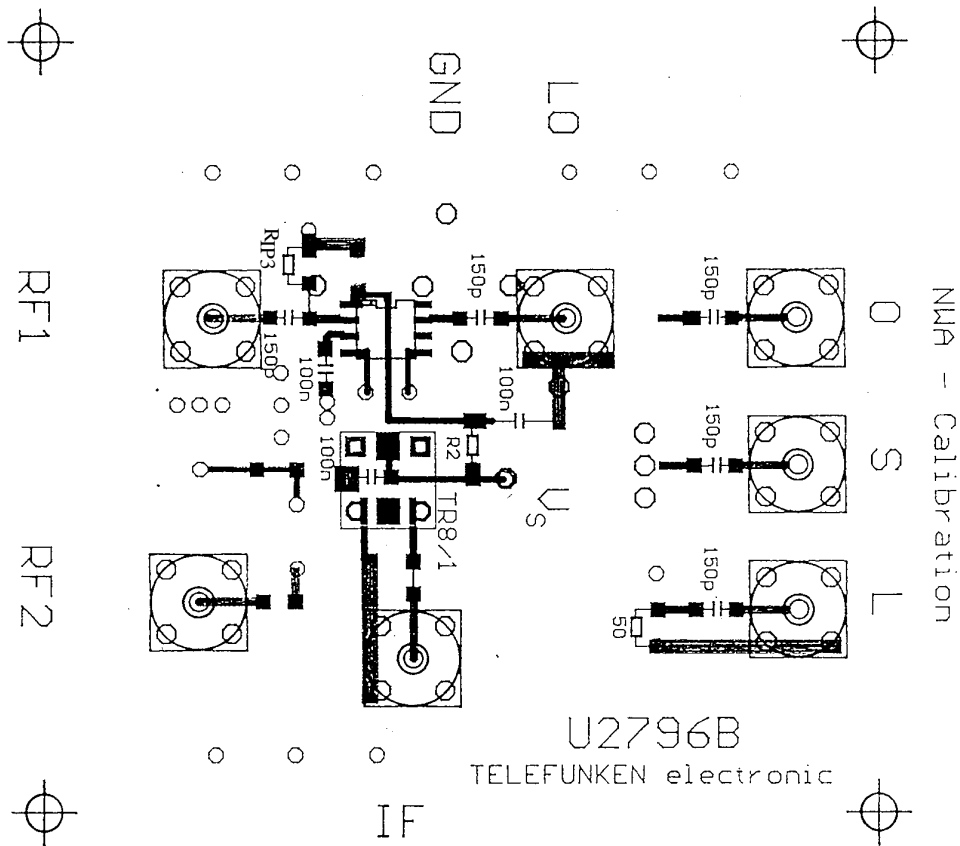


Figure 13.

Evaluation Board

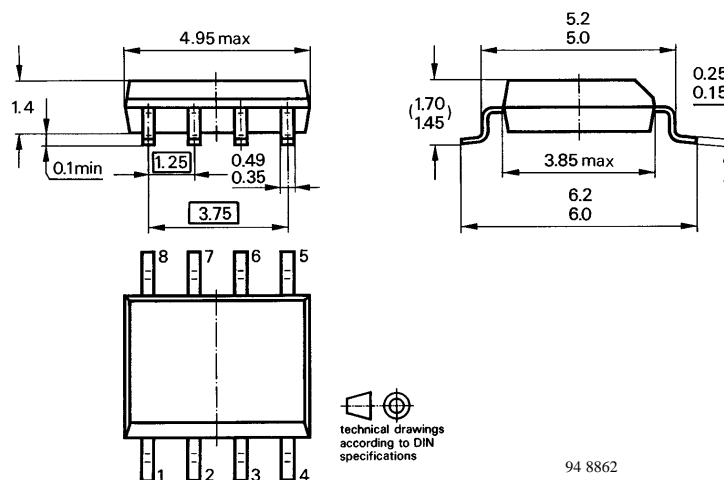


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Figure 14.

Dimensions in mm

SO 8 package



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Ozone Depleting Substances Policy Statement

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1. Meet all present and future national and international statutory requirements.
2. Regularly and continuously improve the performance of our products, processes, distribution and operating systems with respect to their impact on the health and safety of our employees and the public, as well as their impact on the environment.

It is particular concern to control or eliminate releases of those substances into the atmosphere which are known as ozone depleting substances (ODSs).

The Montreal Protocol (1987) and its London Amendments (1990) intend to severely restrict the use of ODSs and forbid their use within the next ten years. Various national and international initiatives are pressing for an earlier ban on these substances.

TEMIC TELEFUNKEN microelectronic GmbH semiconductor division has been able to use its policy of continuous improvements to eliminate the use of ODSs listed in the following documents.

1. Annex A, B and list of transitional substances of the Montreal Protocol and the London Amendments respectively
2. Class I and II ozone depleting substances in the Clean Air Act Amendments of 1990 by the Environmental Protection Agency (EPA) in the USA
3. Council Decision 88/540/EEC and 91/690/EEC Annex A, B and C (transitional substances) respectively.

TEMIC can certify that our semiconductors are not manufactured with ozone depleting substances and do not contain such substances.

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