

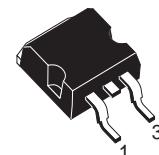


# STB130NH02L

## N-CHANNEL 24V - 0.0034 Ω - 120A D<sup>2</sup>PAK STripFET™ III POWER MOSFET FOR DC-DC CONVERSION

TYPE	V <sub>DSS</sub>	R <sub>D(on)</sub>	I <sub>D</sub>
STB130NH02L	24 V	< 0.0044 Ω	90 A(2)

- TYPICAL R<sub>D(on)</sub> = 0.0034 Ω @ 10 V
- TYPICAL R<sub>D(on)</sub> = 0.005 Ω @ 5 V
- R<sub>D(on)</sub> \* Q<sub>g</sub> INDUSTRY's BENCHMARK
- CONDUCTION LOSSES REDUCED
- SWITCHING LOSSES REDUCED
- LOW THRESHOLD DEVICE
- SURFACE-MOUNTING D<sup>2</sup>PAK (TO-263)  
POWER PACKAGE IN TUBE (NO SUFFIX) OR  
IN TAPE & REEL (SUFFIX "T4")



D<sup>2</sup>PAK  
TO-263  
(Suffix "T4")

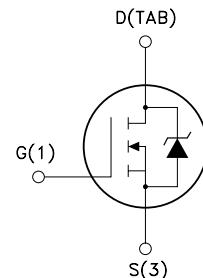
### DESCRIPTION

The STB130NH02L utilizes the latest advanced design rules of ST's proprietary STripFET™ technology. It is ideal in high performance DC-DC converter applications where efficiency is to be achieved at very high output currents.

### APPLICATIONS

- SYNCHRONOUS RECTIFICATIONS FOR TELECOM AND COMPUTER
- OR-ING DIODE

### INTERNAL SCHEMATIC DIAGRAM



SC07580

### Ordering Information

SALES TYPE	MARKING	PACKAGE	PACKAGING
STB130NH02LT4	B130NH02L	TO-263	TAPE & REEL

### ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
V <sub>spike(1)</sub>	Drain-source Voltage Rating	30	V
V <sub>DS</sub>	Drain-source Voltage (V <sub>GS</sub> = 0)	24	V
V <sub>DGR</sub>	Drain-gate Voltage (R <sub>GS</sub> = 20 kΩ)	24	V
V <sub>GS</sub>	Gate-source Voltage	± 20	V
I <sub>D(2)</sub>	Drain Current (continuous) at T <sub>C</sub> = 25°C	90	A
I <sub>D(2)</sub>	Drain Current (continuous) at T <sub>C</sub> = 100°C	90	A
I <sub>DM(3)</sub>	Drain Current (pulsed)	360	A
P <sub>tot</sub>	Total Dissipation at T <sub>C</sub> = 25°C	150	W
	Derating Factor	1	W/°C
E <sub>AS</sub> (4)	Single Pulse Avalanche Energy	900	mJ
T <sub>stg</sub>	Storage Temperature	-55 to 175	°C
T <sub>j</sub>	Max. Operating Junction Temperature		

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### THERMAL DATA

R <sub>thj-case</sub> R <sub>thj-amb</sub> T <sub>I</sub>	Thermal Resistance Junction-case Thermal Resistance Junction-ambient Maximum Lead Temperature For Soldering Purpose	Max Max	1.0 62.5 300	°C/W °C/W °C
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### ELECTRICAL CHARACTERISTICS (T<sub>CASE</sub> = 25 °C UNLESS OTHERWISE SPECIFIED)

OFF

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
V <sub>(BR)DSS</sub>	Drain-source Breakdown Voltage	I <sub>D</sub> = 25 mA, V <sub>GS</sub> = 0	24			V
I <sub>DSS</sub>	Zero Gate Voltage Drain Current (V <sub>GS</sub> = 0)	V <sub>DS</sub> = 20 V V <sub>DS</sub> = 20 V T <sub>C</sub> = 125°C			1 10	μA μA
I <sub>GSS</sub>	Gate-body Leakage Current (V <sub>DS</sub> = 0)	V <sub>GS</sub> = ± 20 V			±100	nA

ON (5)

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
V <sub>GS(th)</sub>	Gate Threshold Voltage	V <sub>DS</sub> = V <sub>GS</sub> I <sub>D</sub> = 250 μA	1			V
R <sub>DS(on)</sub>	Static Drain-source On Resistance	V <sub>GS</sub> = 10 V I <sub>D</sub> = 45 A V <sub>GS</sub> = 5 V I <sub>D</sub> = 22.5 A		0.0034 0.005	0.0044 0.008	Ω Ω

### DYNAMIC

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
g <sub>fs</sub> (5)	Forward Transconductance	V <sub>DS</sub> = 10 V I <sub>D</sub> = 45 A		55		S
C <sub>iss</sub> C <sub>oss</sub> C <sub>rss</sub>	Input Capacitance Output Capacitance Reverse Transfer Capacitance	V <sub>DS</sub> = 15V f = 1 MHz V <sub>GS</sub> = 0		4450 1126 141		pF pF pF
R <sub>G</sub>	Gate Input Resistance	f = 1 MHz Gate DC Bias = 0 Test Signal Level = 20 mV Open Drain		1.6		Ω

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### ELECTRICAL CHARACTERISTICS (continued)

#### SWITCHING ON

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
$t_{d(on)}$ $t_r$	Turn-on Delay Time Rise Time	$V_{DD} = 10 \text{ V}$ $I_D = 45 \text{ A}$ $R_G = 4.7 \Omega$ $V_{GS} = 10 \text{ V}$ (Resistive Load, Figure 3)		14 224		ns ns
$Q_g$ $Q_{gs}$ $Q_{gd}$	Total Gate Charge Gate-Source Charge Gate-Drain Charge	$V_{DD} = 10 \text{ V}$ $I_D = 90 \text{ A}$ $V_{GS} = 10 \text{ V}$		69 13 9	93	nC nC nC
$Q_{oss}^{(6)}$	Output Charge	$V_{DS} = 16 \text{ V}$ $V_{GS} = 0 \text{ V}$		27		nC
$Q_{gls}^{(7)}$	Third-quadrant Gate Charge	$V_{DS} < 0 \text{ V}$ $V_{GS} = 10 \text{ V}$		64		nC

#### SWITCHING OFF

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
$t_{d(off)}$ $t_f$	Turn-off Delay Time Fall Time	$V_{DD} = 10 \text{ V}$ $I_D = 45 \text{ A}$ $R_G = 4.7 \Omega$ , $V_{GS} = 10 \text{ V}$ (Resistive Load, Figure 3)		69 40	54	ns ns

#### SOURCE DRAIN DIODE

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
$I_{SD}$ $I_{SDM}$	Source-drain Current Source-drain Current (pulsed)				90 360	A A
$V_{SD}^{(5)}$	Forward On Voltage	$I_{SD} = 45 \text{ A}$ $V_{GS} = 0$			1.3	V
$t_{rr}$ $Q_{rr}$ $I_{RRM}$	Reverse Recovery Time Reverse Recovery Charge Reverse Recovery Current	$I_{SD} = 90 \text{ A}$ $di/dt = 100 \text{ A}/\mu\text{s}$ $V_{DD} = 15 \text{ V}$ $T_j = 150^\circ\text{C}$ (see test circuit, Figure 5)		47 58 2.5		ns nC A

(1) Guaranteed when external  $R_G=4.7 \Omega$  and  $t_f < t_{fmax}$ .

(2) Value limited by wire bonding

(3) Pulse width limited by safe operating area.

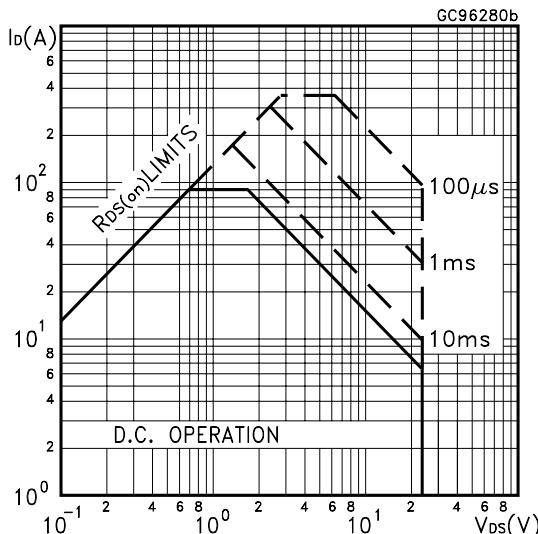
(4) Starting  $T_j = 25^\circ\text{C}$ ,  $I_D = 45 \text{ A}$ ,  $V_{DD} = 10 \text{ V}$ .

(5) Pulsed: Pulse duration = 300  $\mu\text{s}$ , duty cycle 1.5 %.

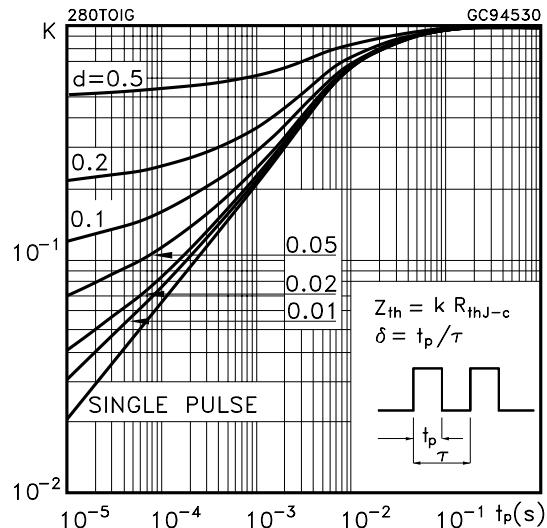
(6)  $Q_{oss} = C_{oss} * \Delta V_{in}$ ,  $C_{oss} = C_{gd} + C_{ds}$ . See Appendix A

(7) Gate charge for synchronous operation

#### Safe Operating Area

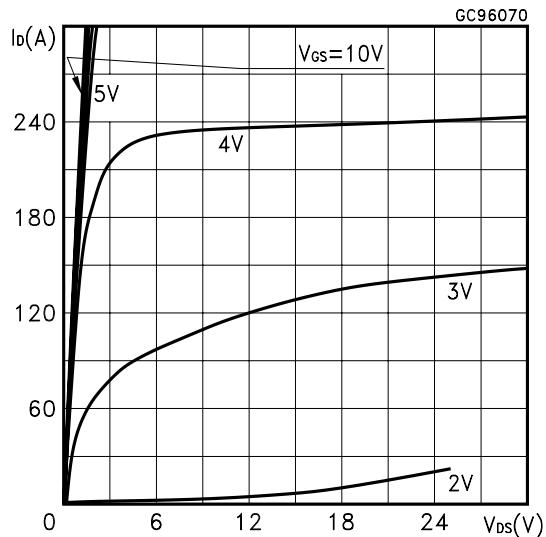


#### Thermal Impedance

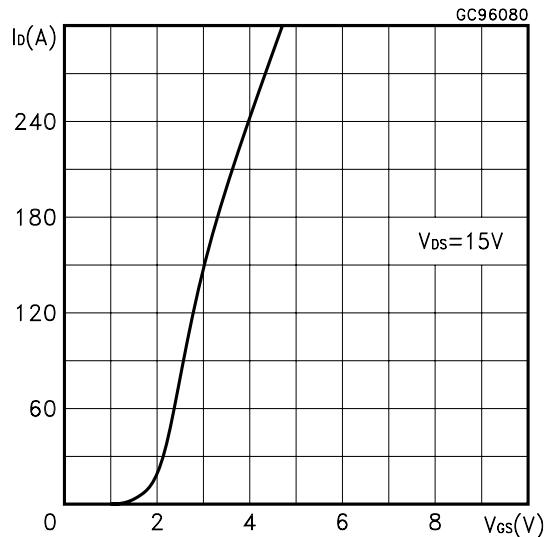


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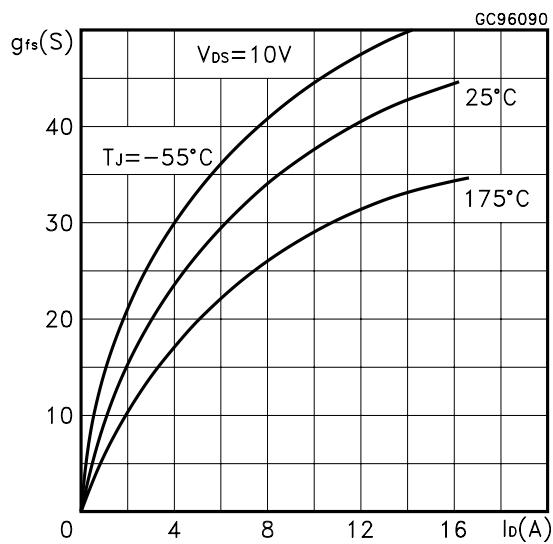
Output Characteristics



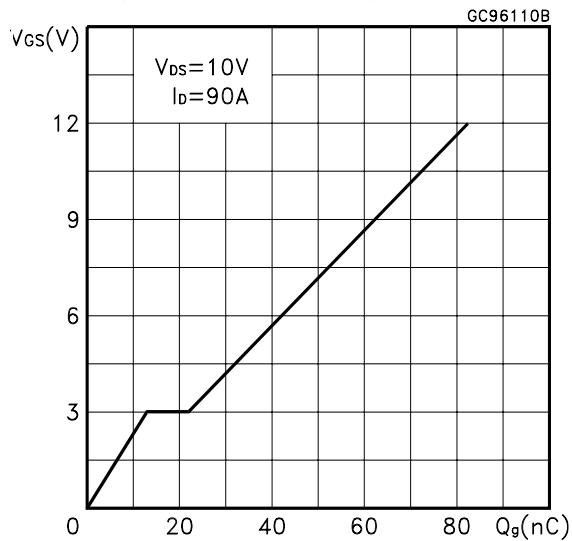
Transfer Characteristics



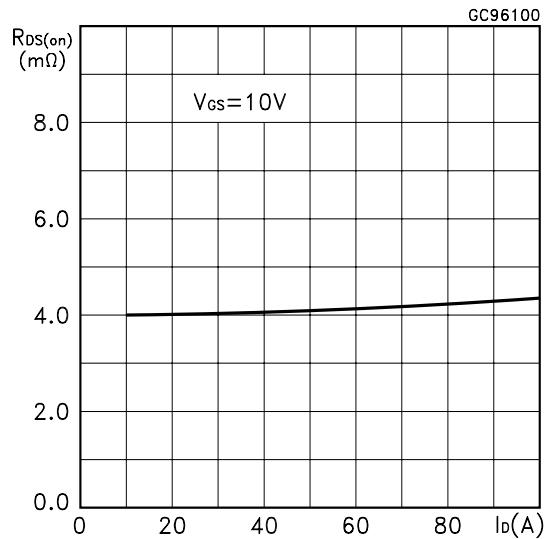
Transconductance



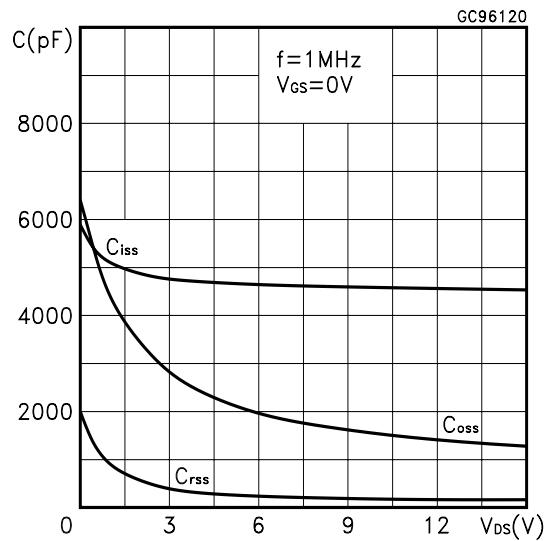
Gate Charge vs Gate-source Voltage



Static Drain-source On Resistance

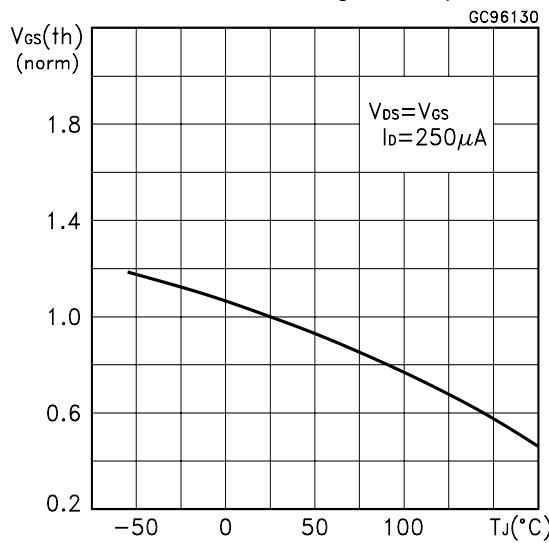


Capacitance Variations

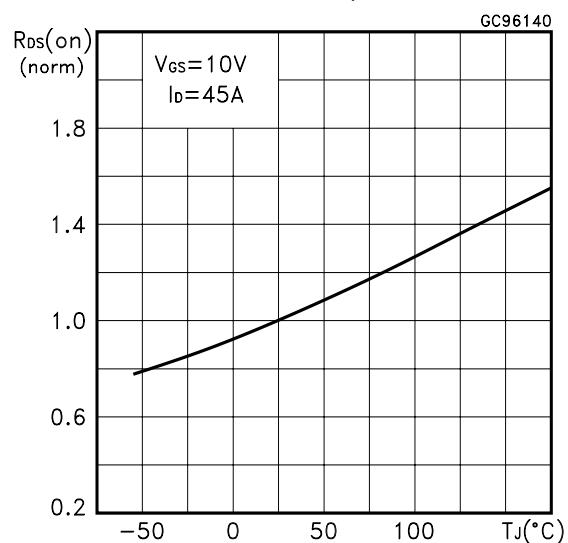


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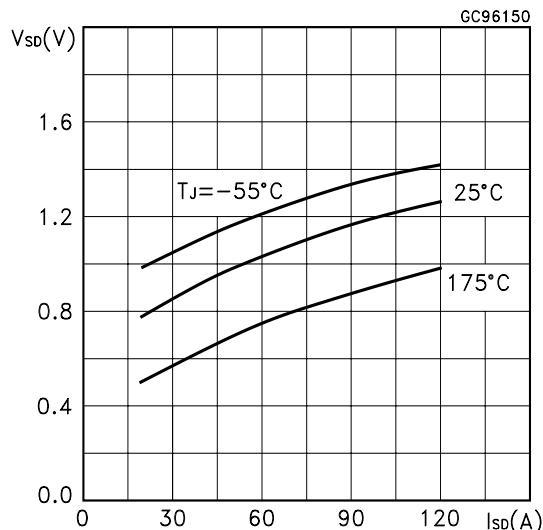
Normalized Gate Threshold Voltage vs Temperature



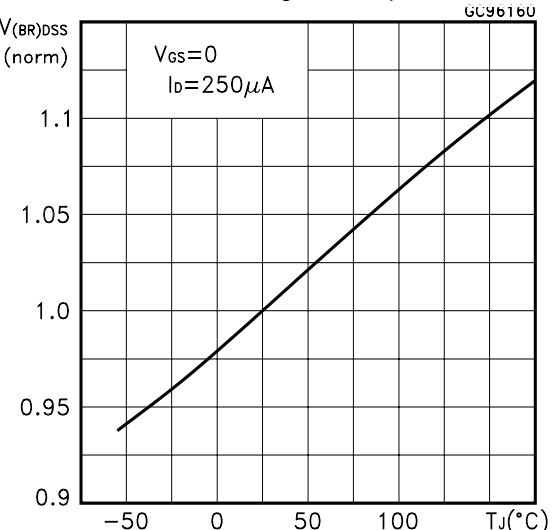
Normalized on Resistance vs Temperature



Source-drain Diode Forward Characteristics

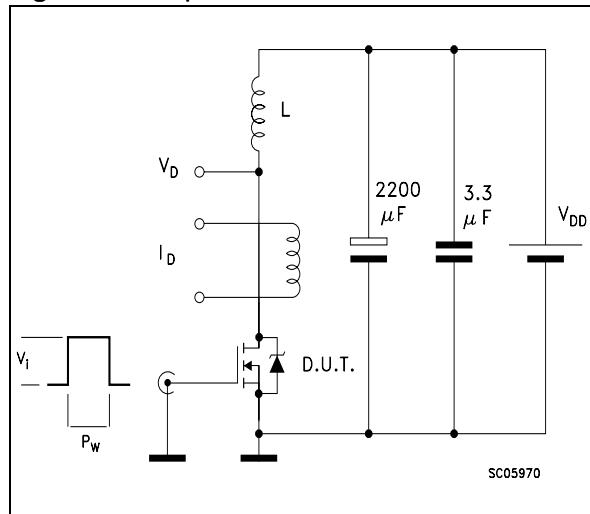


Normalized Breakdown Voltage vs Temperature

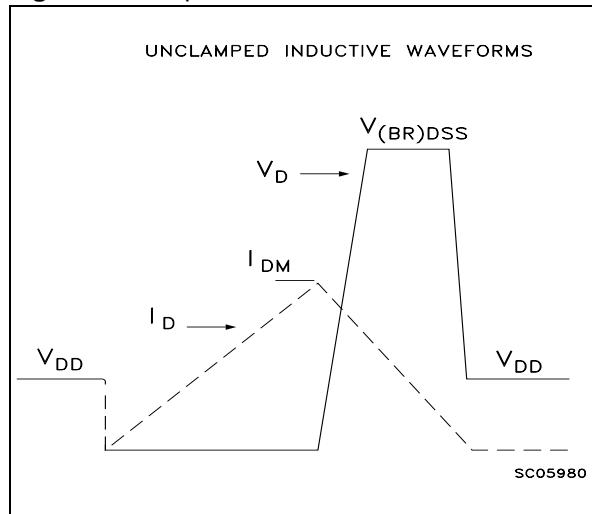


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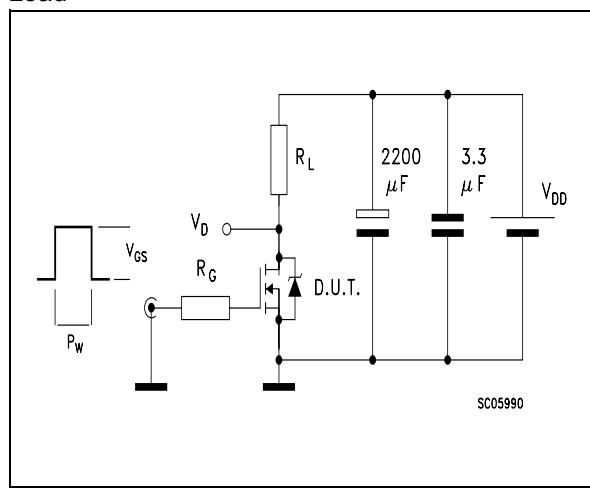
**Fig. 1: Unclamped Inductive Load Test Circuit**



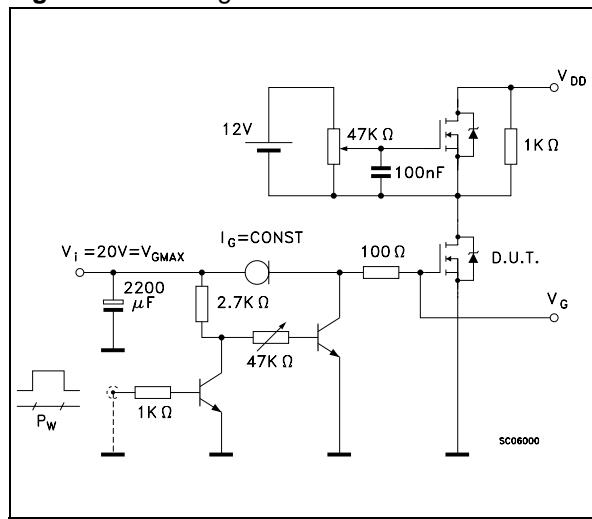
**Fig. 2: Unclamped Inductive Waveform**



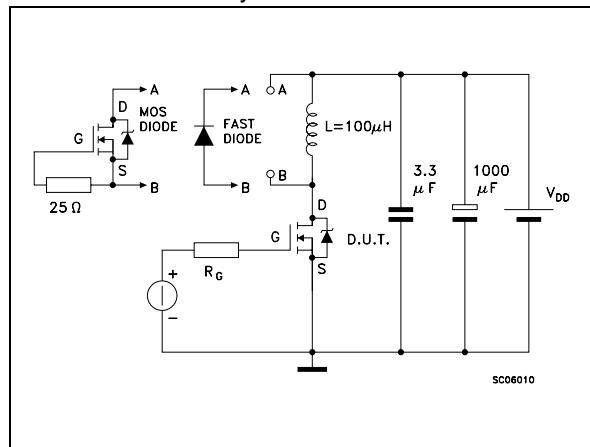
**Fig. 3: Switching Times Test Circuits For Resistive Load**



**Fig. 4: Gate Charge test Circuit**



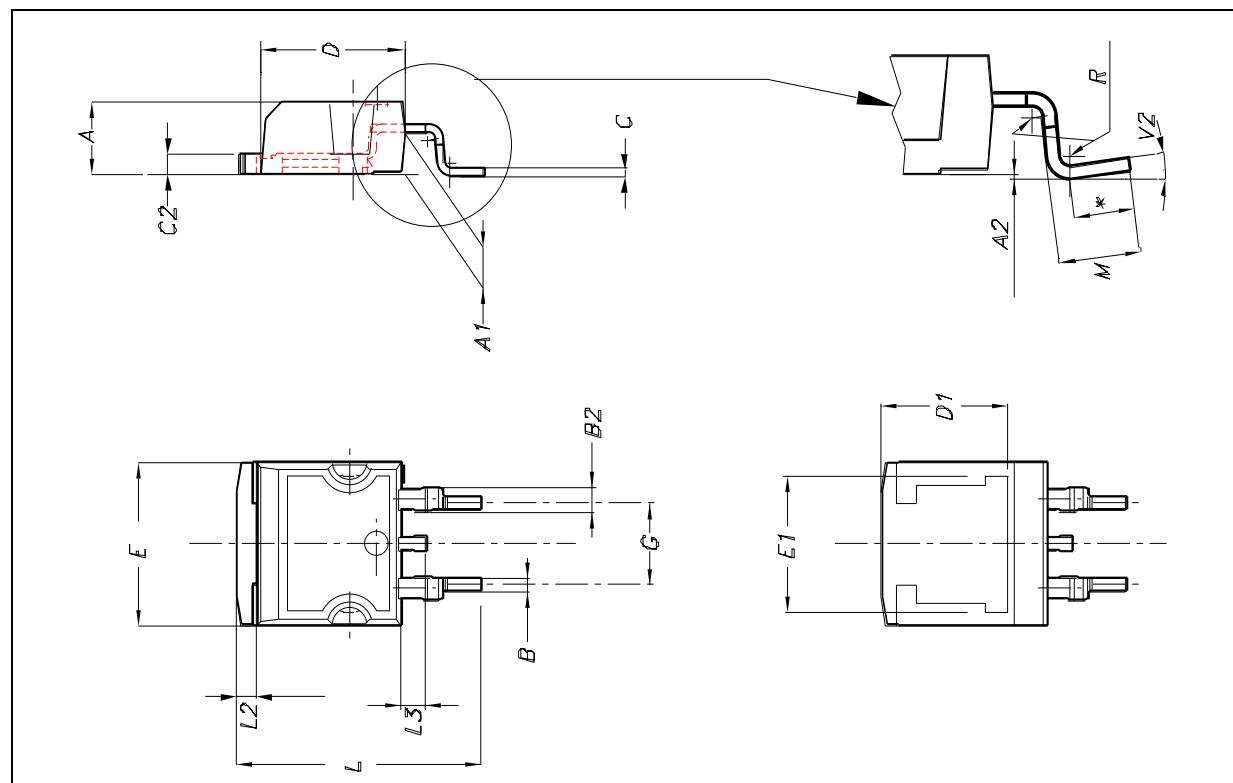
**Fig. 5: Test Circuit For Inductive Load Switching And Diode Recovery Times**

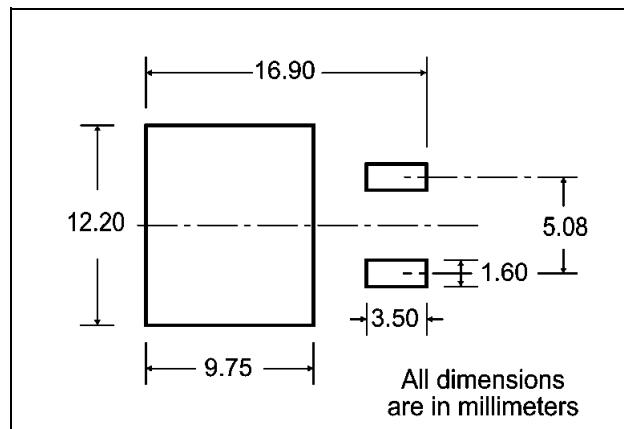
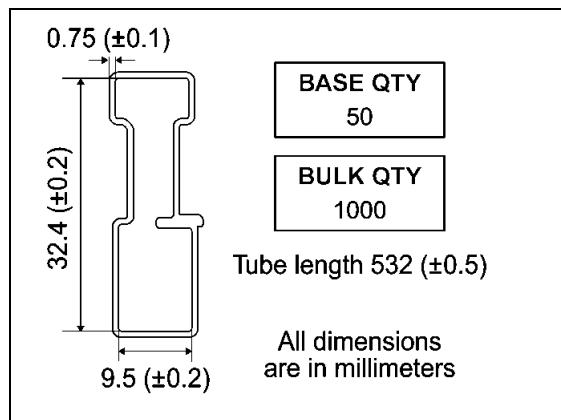
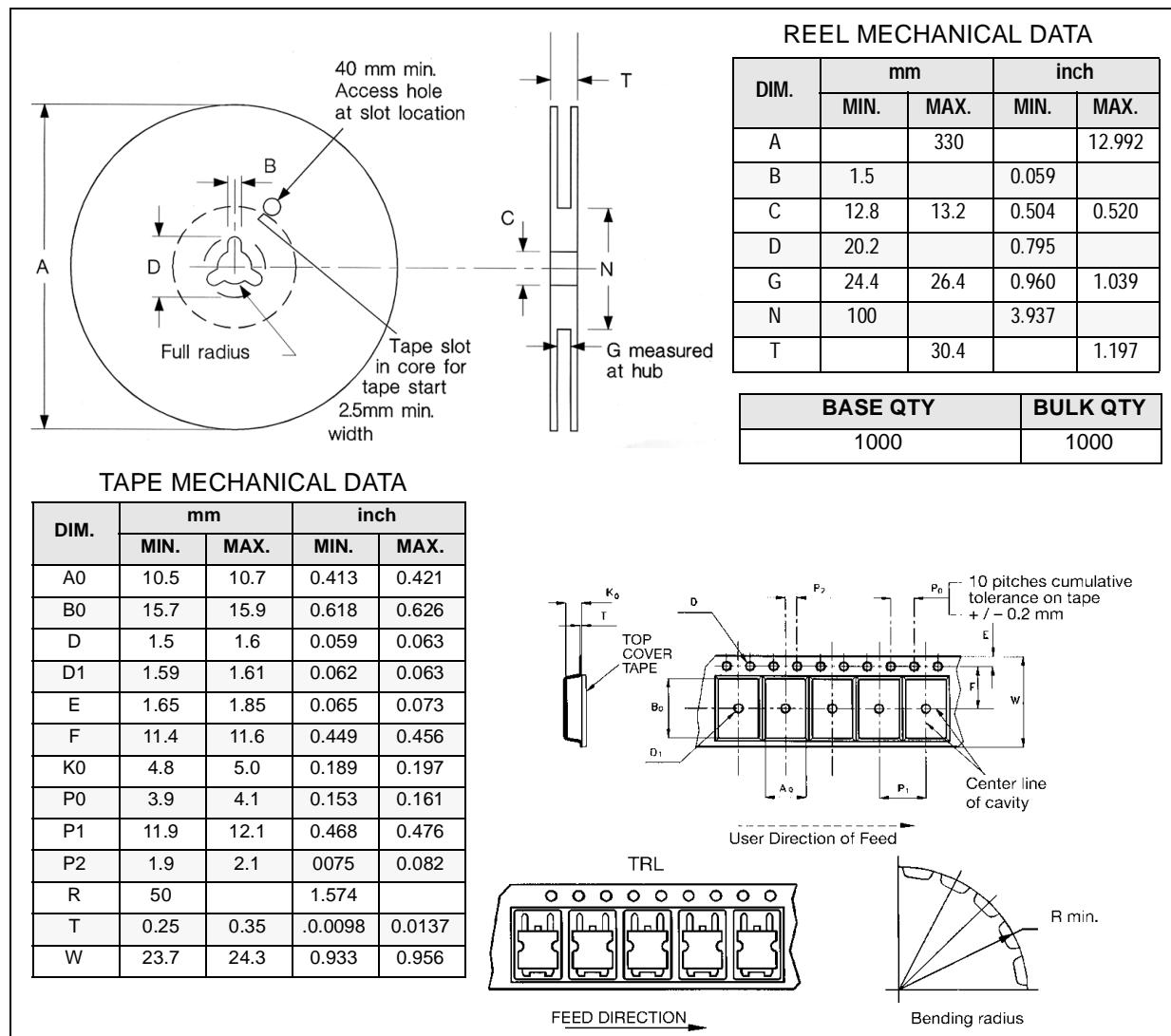


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### D<sup>2</sup>PAK MECHANICAL DATA

DIM.	mm.			inch.		
	MIN.	TYP.	MAX.	MIN.	TYP.	TYP.
<b>A</b>	4.4		4.6	0.173		0.181
<b>A1</b>	2.49		2.69	0.098		0.106
<b>A2</b>	0.03		0.23	0.001		0.009
<b>B</b>	0.7		0.93	0.028		0.037
<b>B2</b>	1.14		1.7	0.045		0.067
<b>C</b>	0.45		0.6	0.018		0.024
<b>C2</b>	1.21		1.36	0.048		0.054
<b>D</b>	8.95		9.35	0.352		0.368
<b>D1</b>		8			0.315	
<b>E</b>	10		10.4	0.394		0.409
<b>E1</b>		8.5			0.334	
<b>G</b>	4.88		5.28	0.192		0.208
<b>L</b>	15		15.85	0.591		0.624
<b>L2</b>	1.27		1.4	0.050		0.055
<b>L3</b>	1.4		1.75	0.055		0.069
<b>M</b>	2.4		3.2	0.094		0.126
<b>R</b>		0.4			0.015	
<b>V2</b>	0°		8°	0°		8°

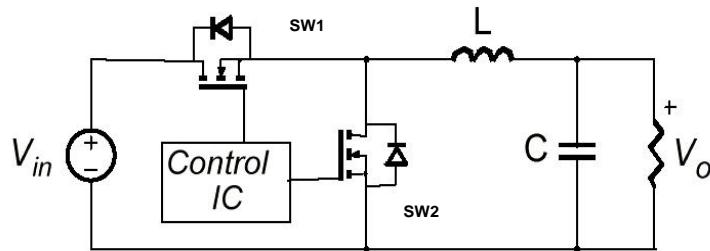


**D<sup>2</sup>PAK FOOTPRINT****TUBE SHIPMENT (no suffix)\*****TAPE AND REEL SHIPMENT (suffix "T4")\***

\* on sales type

## APPENDIX A

### Buck Converter: Power Losses Estimation



The power losses associated with the FETs in a Synchronous Buck converter can be estimated using the equations shown in the table below. The formulas give a good approximation, for the sake of performance comparison, of how different pairs of devices affect the converter efficiency. However a very important parameter, the working temperature, is not considered. The real device behavior is really dependent on how the heat generated inside the devices is removed to allow for a safer working junction temperature.

The low side (**SW2**) device requires:

- Very low  $R_{DS(on)}$  to reduce conduction losses
- Small  $Q_{gls}$  to reduce the gate charge losses
- Small  $C_{oss}$  to reduce losses due to output capacitance
- Small  $Q_{rr}$  to reduce losses on  $SW_1$  during its turn-on
- The  $C_{gd}/C_{gs}$  ratio lower than  $V_{th}/V_{gg}$  ratio especially with low drain to source voltage to avoid the cross conduction phenomenon;

The high side (**SW1**) device requires:

- Small  $R_g$  and  $L_s$  to allow higher gate current peak and to limit the voltage feedback on the gate
- Small  $Q_g$  to have a faster commutation and to reduce gate charge losses
- Low  $R_{DS(on)}$  to reduce the conduction losses.

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		<b>High Side Switch (SW1)</b>	<b>Low Side Switch (SW2)</b>
$P_{\text{conduction}}$		$R_{DS(\text{on})\text{SW1}} * I_L^2 * d$	$R_{DS(\text{on})\text{SW2}} * I_L^2 * (1-d)$
$P_{\text{switching}}$		$V_{in} * (Q_{gsth(\text{SW1})} + Q_{gd(\text{SW1})}) * f * \frac{I_L}{I_g}$	Zero Voltage Switching
$P_{\text{diode}}$	Recovery	Not Applicable	$^1 V_{in} * Q_{rr(\text{SW2})} * f$
	Conduction	Not Applicable	$V_{f(\text{SW2})} * I_L * t_{deadtime} * f$
$P_{\text{gate}(Q_G)}$		$Q_{g(\text{SW1})} * V_{gg} * f$	$Q_{gls(\text{SW2})} * V_{gg} * f$
$P_{Qoss}$		$\frac{V_{in} * Q_{oss(\text{SW1})} * f}{2}$	$\frac{V_{in} * Q_{oss(\text{SW2})} * f}{2}$

<b>Parameter</b>	<b>Meaning</b>
$d$	Duty-cycle
$Q_{gsth}$	Post threshold gate charge
$Q_{gls}$	Third quadrant gate charge
$P_{\text{conduction}}$	On state losses
$P_{\text{switching}}$	On-off transition losses
$P_{\text{diode}}$	Conduction and reverse recovery diode losses
$P_{\text{gate}}$	Gate drive losses
$P_{Qoss}$	Output capacitance losses

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<sup>1</sup> Dissipated by SW1 during turn-on

## **STB130NH02L**

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