

## Preliminary

**RF2460** 

PCS CDMA LOW NOISE AMPLIFIER/MIXER 1500MHZ TO 2200MHZ DOWNCONVERTER

### Typical Applications

- CDMA PCS Handsets
- GPS Receiver
- W-CDMA Handsets

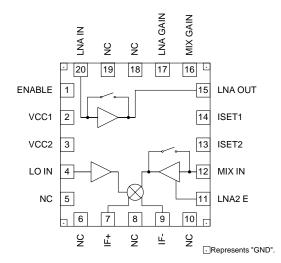
- General Purpose Downconverter
- Commercial and Consumer Systems
- Portable Battery-Powered Equipment

### **Product Description**

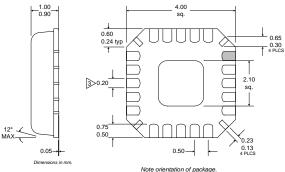
The RF2460 is a receiver front-end designed for the receive section of PCS CDMA and W-CDMA applications. It is designed to amplify and downconvert RF signals while providing 29dB of stepped gain control range and features digital control of LNA gain, mixer gain, and power down mode. A further feature of the chip is adjustable IIP3 of the LNA and mixer using an off-chip current setting resistor. Noise Figure, IP3, and other specs are designed to be compatible with the IS-98B for CDMA PCS communications. The IC is manufactured on a SiGeHBT process and packaged in a 20-pin leadless chip carrier with an exposed die flag.

Optimum Technology Matching® Applied

☐ Si BJT ☐ GaAs HBT ☐ GaAs MESFET☐ Si Bi-CMOS ▼ SiGe HBT ☐ Si CMOS



Functional Block Diagram



#### NOTES:

- Shaded lead is Pin 1.
- 2 Pin 1 identifier must exist on top surface of package by identification mark or feature on the package body. Exact shape and size is optional.
- Dimension applies to plated terminal: to be measured between 0.02 mm and 0.25 mm from terminal end.
- and 0.25 mm from terminal end.

  4 Package Warpage: 0.05 mm max.
- 5 Die Thickness Allowable: 0.305 mm max.

Package Style: LCC, 20-Pin, 4x4

### **Features**

- Complete Receiver Front-End
- Stepped LNA/Mixer Gain Control
- Adjustable LNA/Mixer Bias Current
- 24dB Gain and 2.2dB Noise Figure at Maximum Cascade Gain

#### Ordering Information

RF2460 PCS CDMA Low Noise Amplifier/Mixer 1500MHz to

2200 MHz Downconverter

RF2460 PCBA Fully Assembled Evaluation Board

 RF Micro Devices, Inc.
 Tel (336) 664 1233

 7628 Thorndike Road
 Fax (336) 664 0454

 Greensboro, NC 27409, USA
 http://www.rfmd.com

### **Absolute Maximum Ratings**

	•	
Parameter	Rating	Unit
Supply Voltage	-0.5 to +5.0	$V_{DC}$
Input LO and RF Levels	+6	dBm
Operating Ambient Temperature	-40 to +85	°C
Storage Temperature	-40 to +150	°C



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Dozomotor	Specification			Unit	Condition		
Parameter	Min. Typ.		Max.				
Overall					T=25°C, V <sub>CC</sub> =2.75V, RF=1.96GHz,		
RF Frequency Range		1500 to 2200		MHz	LO=2170MHz@-7dBm, IF=210MHz		
LO Frequency Range		1200 to 2600		MHz			
IF Frequency Range		0.1 to 250		MHz			
Bias Current		2.5	2.8	mA	LNA, mixer and preamp for bias circuitry.		
LNA		2.0	2.0	IIIA	EIVA, Illixer and preamp for bias circuity.		
Gain	13.5	15.0		dB			
Noise Figure	13.3	1.4	1.8	dB			
Input IP3	+6.0	+7.0	1.0	dBm	IIP3 is adjustable (see plots for setting).		
input ii o	10.0	17.0		d d d d	ISET1 (pin 14) external resistor sets current		
Input VSWR			2:1		consumption and performance.		
Output VSWR			2:1				
Current at Input IP3		7	7.5	mA			
LNA Bypass			7.0	1177			
Gain	-6	-5		dB			
Noise Figure		5	5.5	dB			
Input IP3	+23.0	+26.0	0.0	dBm			
Input VSWR	120.0	120.0	2:1	45			
Output VSWR			2:1				
Current		0		mA			
Mixer - High Gain Mode					1kΩ balanced load.		
Gain	10	12		dB			
Noise Figure		6.5	7.5	dB			
Input IP3	+3.0	+4.0		dBm	IIP3 is adjustable (see plots for setting).		
RF to IF Isolation		>45		dB	ISET2 (pin 13) external resistor sets current		
					consumption and performance.		
Input VSWR			2:1				
Output VSWR			2:1				
Current		12	13	mA			
Mixer - Low Gain Mode			-		1kΩ balanced load.		
Gain	0	1.5		dB			
Noise Figure		15	16	dB			
Input IP3	+13.0	+14.0		dBm	IIP3 is adjustable		
RF to IF Isolation		>45		dB	ISET2 (pin 13) external resistor sets current		
					consumption and performance.		
Input VSWR			2:1				
Output VSWR			2:1				
Current		7.5	8.0	mA			

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Daramatar	Specification			Unit	Condition		
Parameter	Min.	Тур.	Max.	Unit	Condition		
GPS - LNA							
Gain		16		dB			
Noise Figure		1.4		dB			
Input IP3		+7.0		dBm	IIP3 is adjustable. ISET1 (pin 14) external resistor sets current consumption and per-		
Current at Input IP3		7		mA	formance.		
GPS - Mixer	+			IIIA	+		
Gain		17		dB			
		6		dВ			
Noise Figure		-5.0		dВm	IID2 in adjustable ISET1 (nin 14) external		
Input IP3		-5.0		авт	IIP3 is adjustable. ISET1 (pin 14) external resistor sets current consumption and performance.		
Current at Input IP3		16		mA			
GPS - Cascaded							
Gain		31		dB			
Noise Figure		2.0		dB			
Input IP3		-1.0		dBm	IIP3 is adjustable. ISET1 (pin 14) external resistor sets current consumption and performance.		
Current at Input IP3		23		mA			
Local Oscillator Input							
Input Level	-10	-7	0	dBm			
LO to RF Isolation		>40		dB	Any gain state.		
LO to LNA Isolation		>60		dB	Any gain state.		
LO Current Buffer		4.5	5.0	mA	I <sub>CC2</sub> when LO signal is present		
Cascade - LNA High/Mixer High					LNA High Gain/Mixer High Gain Assuming 3dB loss of filter		
Gain		24		dB	IF 1, 1kΩ balanced load.		
Noise Figure		2.2		dB	,		
Input IP3		-8.0		dBm	Single sideband.		
Total Current		26		mA			
Cascade -					LNA High Gain/Mixer Low Gain		
LNA High/Mixer Low					Assuming 3dB loss of filter		
Gain		13.5		dB	IF 1, 1kΩ balanced load.		
Noise Figure		5.3		dB			
Input IP3		+1.0		dBm	Single sideband.		
Total Current		21		mA			
Cascade -					LNA Low Gain/Mixer High Gain		
LNA Low/Mixer High					Assuming 3dB loss of filter		
Gain		4		dB	IF 1, 1k $\Omega$ balanced load.		
Noise Figure		14.5		dB			
Input IP3		+12.0		dB	Single sideband.		
Total Current	<u> </u>	19		mA			
Cascade -					LNA Low Gain/Mixer Low Gain		
LNA Low/Mixer Low					Assuming 3dB loss of filter		
Gain		-6.5		dB	IF 1, $1k\Omega$ balanced load.		
Noise Figure		23		dB			
Input IP3		+20.5		dB	Single sideband.		
Total Current		14		mA			
Power Supply							
Voltage	2.7	3.0	3.3	V			

1 1	ENABLE	Power down pin. A logic "low" turns the part off. A logic "high" (>1.6V)	
2		turns the part on.	
	VCC1	Supply Voltage for the LNA, mixer, bias, and logic circuitry. External RF and IF bypassing is required. The trace length between the pin and the bypass capacitors should be minimized. The ground side of the bypass capacitors should connect immediately to ground plane.	See pin 20.
3	VCC2	Supply Voltage for the LO buffer amplifier. External RF and IF bypassing is required. The trace length between the pin and the bypass capacitors should be minimized. The ground side of the bypass capacitors should connect immediately to ground plane.	
4	LO IN	Mixer LO Input Pin.	
5	NC	No connection. For isolation purposes, this pin is connected to the ground plane.	
6	NC	No connection. For isolation purposes, this pin is connected to the ground plane.	
7	IF+	CDMA IF Output pin. This is a balanced output. The internal circuitry, in conjunction with an external matching/bias inductor to $V_{CC}$ , sets the operating impedance. This inductor is typically incorporated in the matching network between the output and IF filter. The part is designed to drive a $1\mathrm{k}\Omega$ load. Because this pin is biased to $V_{CC}$ , a DC blocking capacitor must be used if the IF filter input has a DC path to ground. See Application Schematic.	1.2 pF 1.2 pF
8	NC	No connection. For isolation purposes, this pin is connected to the ground plane.	
9	IF-	Same as pin 7, except complementary output.	See pin 6.
10	NC	No connection. For isolation purposes, this pin is connected to the ground plane.	
11	LNA2 E	Emitter for LNA2. Increasing the inductance on this pin will reduce the mixer gain, increase IP3 and noise figure.	
12	MIX IN	Mixer RF Input Pin. This pin is internally DC biased and should be DC blocked if connected to a device with DC present. External matching network sets RF and IF impedance for optimum performance.	MIX IN O
13	ISET2	This pin is used to set the bias current and IIP3 of the mixer amplifier using a resistor to ground. See plots for values and current settings.	
14	ISET1	This pin is used to set the bias current and IIP3 of the LNA amplifier using a resistor to ground. See plots for values and current settings.	
15 L	_NA OUT	LNA output pin. Open collector.	See pin 20.
16 N	MIX GAIN	CMOS compatible signal controlling mixer gain mode. Setting this signal high places the mixer in the high gain mode. Setting this signal low places the mixer in low gain mode by bypassing and shutting off the mixer buffer amplifier current.	MIX GAIN O——
17 L	NA GAIN	CMOS compatible signal controlling LNA gain mode. Setting this signal high places the LNA in the high gain mode. Setting this signal low bypasses the LNA and shuts off the LNA bias current.	LNA GAIN O———
18	NC	No connection. For isolation purposes, this pin is connected to the ground plane.	
19	NC	No connection. For isolation purposes, this pin is connected to the ground plane.	

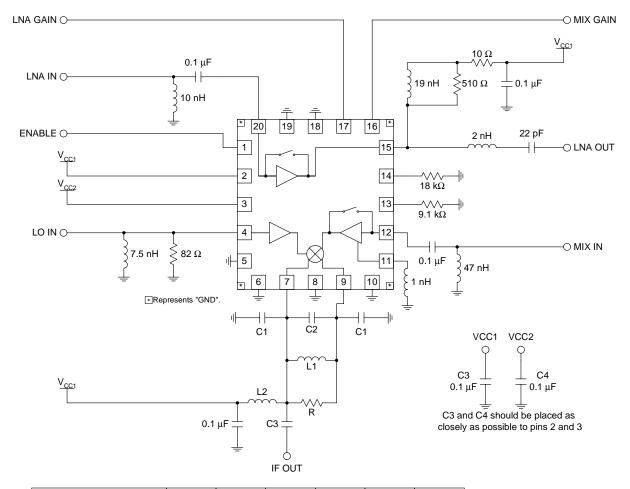
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# Preliminary

Pin	Function	Description	Interface Schematic
20	LNA IN	RF Input pin. This pin is internally matched for optimum noise figure from a $50\Omega$ source.	LNA IN O
Pkg Base	GND	Ground connection. The backside of the package should be soldered to a top side ground pad which is connected to the ground plane with multiple vias.	

## Application Schematic - US PCS



	C1 (pF)	C2 (pF)	C3 (pF)	L1 (nH)	L2 (nH)	R (Ω)
US PCS, IF = 210 MHz	4	3	6	82	110	4.7 k
Korean PCS, IF = 220 MHz	3.6	2	7	82	120	4.7 k
GPS, IF = 184 MHz	4	3	5	150	82	3 k
US PCS, IF = 184 MHz	8	3	6	82	110	4.7 k

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FRONT-ENDS

## Output Interface Network of the Mixer

L1, C1, C2, and R form a current combiner which performs a differential to single-ended conversion at the IF frequency and sets the output impedance. In most cases, the resonance frequency is independent of R and can be set according to the following equation:

$$f_{IF} = \frac{1}{2\pi \sqrt{\frac{L1}{2}(C_1 + 2C_2 + C_{EQ})}}$$

Where  $C_{EQ}$  is the equivalent stray capacitance and capacitance looking into pins 7 and 9. An average value to use for  $C_{EQ}$  is 2.5 pF.

R can then be used to set the output impedance according to the following equation:

$$R = \left(\frac{1}{4 \cdot R_{OUT}} - \frac{1}{R_P}\right)^{-1}$$

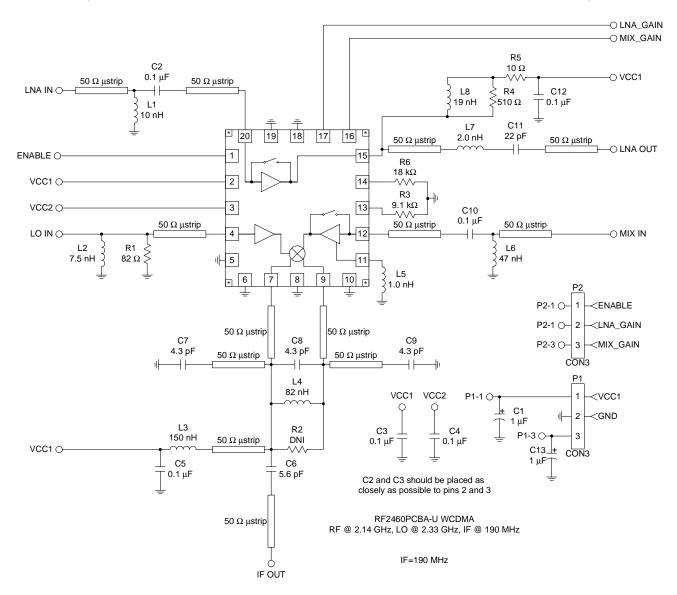
where  $R_{OUT}$  is the desired output impedance and  $R_P$  is the parasitic equivalent parallel resistance of L1.

 $C_2$  should first be set to 0 and C1 should be chosen as high as possible (suggested less than 20pF), while maintaining an  $R_P$  of L1 that allows for the desired  $R_{OUT}$ . If the self-resonant frequencies of the selected C1 produce unsatisfactory linearity performance, their values may be reduced and compensated for by including C2 capacitor with a value chosen to maintain the desired  $F_{IF}$  frequency.

L2 and C3 serve dual purposes. L2 serves as an output bias choke, and C3 serves as a series DC block.

In addition, L2 and C3 may be chosen to form an impedance matching network if the input impedance of the IF filter is not equal to R<sub>OUT</sub>. Otherwise, L2 is chosen to be large (suggested 120nH) and C3 is chosen to be large (suggested 22nF) if a DC path to ground is present in the IF filter, or omitted if the filter is DC blocked.

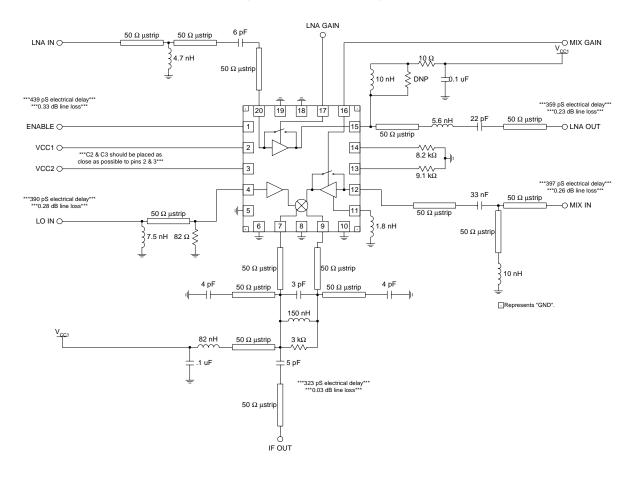
# Application Schematic - W-CDMA (See W-CDMA charts for lab measurements at the end of the data sheet)



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# Application Schematic - GPS RF=1575MHz, IF=184MHz, LO=1759MHz



### **Current Measurement**

To measure only the current of the different circuitry in the evaluation board, use the following procedure.

First, replace the bias choke inductor at the output of the mixer (L3 for US-PCS) with a  $1\Omega$  resistor. The voltage across the resistor will represent the mixer current. Terminate all SMA connections at  $50\Omega$ .

Second, follow the table below.

### CONDITION

	Current (mA)	V <sub>CC1</sub>	V <sub>CC2</sub>	EN	LNA Gain	Mix Gain
I <sub>CC</sub> Total	25.82	1	1	1	1	1
LNA Off	18.77	1	1	1	0	1
Mixer Preamp Off	14.28	1	1	1	0	0
V <sub>CC2</sub> Off	10.05	1	0	1	0	0
Mixer Current	7.72	1	0	1	0	0

Therefore,

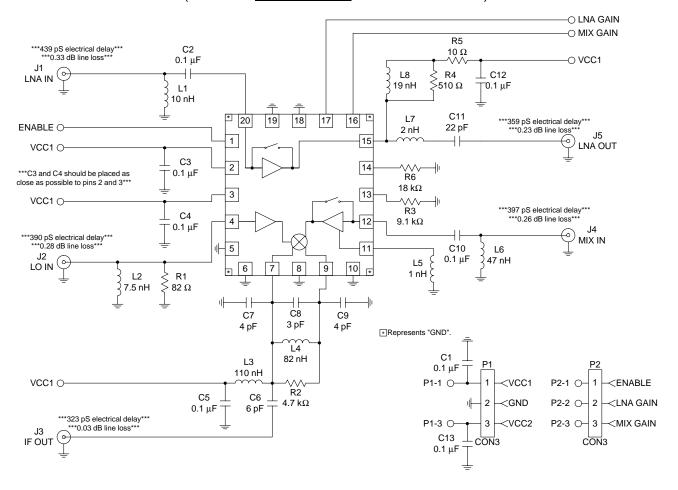
LNA (Bypass) = (Computer Simulation) = 0mA <sup>.</sup> 25.82-18.77 LNA (High Gain) = 7.05 mA Mixer (Preamp) = 18.77-14.28 4.49 mA Mixer = 7.70mA (Measured) Bias = 10.05-7.7 2.35 mA 14.28-10.05 LO Circuitry (V<sub>CC2</sub>) = 4.23 mA 25.82mA

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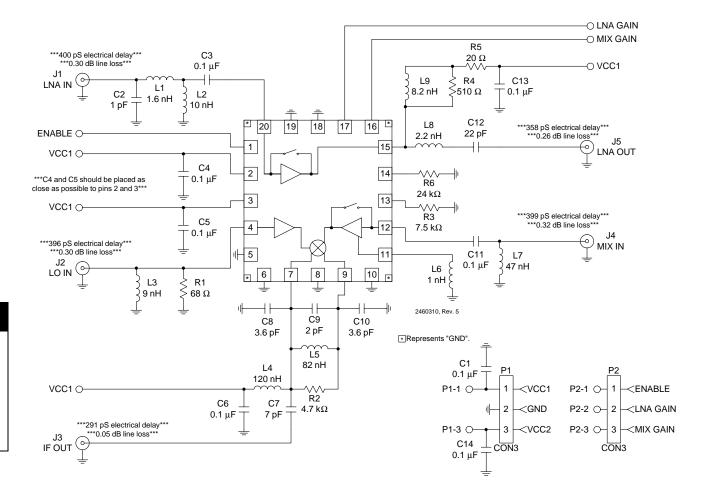
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# Evaluation Board Schematic US-PCS, IF=210MHz

(Download Bill of Materials from www.rfmd.com.)



# Evaluation Board Schematic Korean-PCS, IF=220MHz

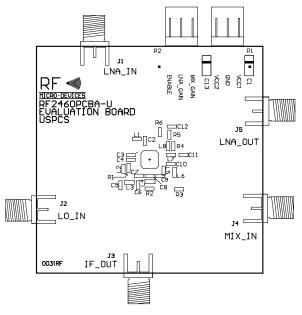


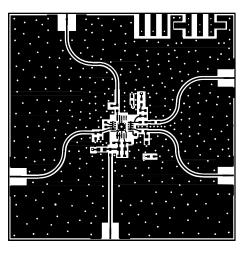
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## Evaluation Board Layout - US PCS Board Size 2.0" x 2.0"

Board Thickness 0.034", Board Material FR-4, Multi-Layer

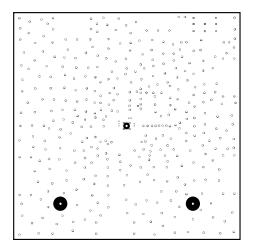




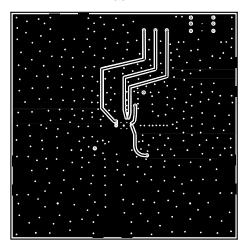


Power Plane 1

Power Plane 2

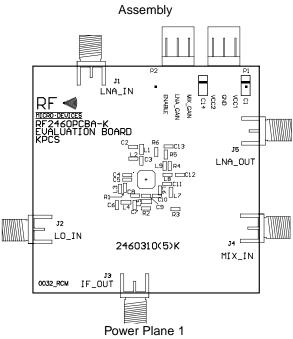


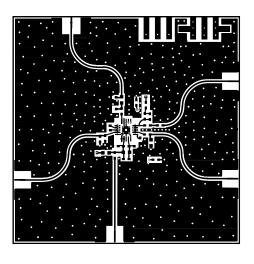


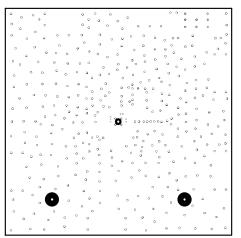


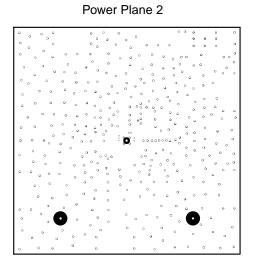
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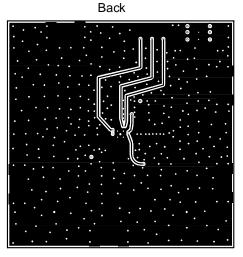
# **Evaluation Board Layout - Korean PCS**



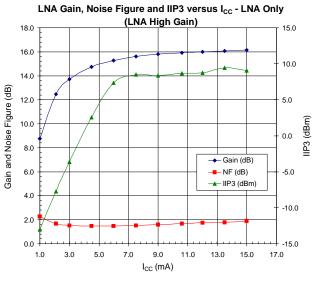


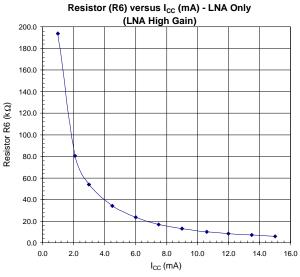


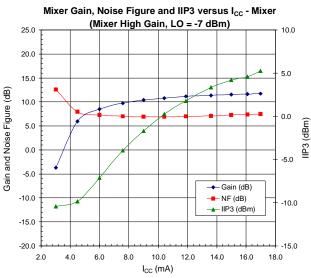


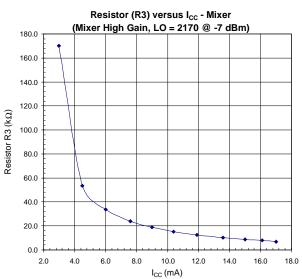


#### **US-PCS**









Special Instructions (Board loss, taking into consideration description in the schematic)

LNA

 $V_{CC1}=V_{CC2}=Enable=2.75V$ ; Mix Gain=0.0V

To measure I<sub>CC</sub> LNA only:

LNA Gain was switched between 0V and 2.75V, and record the delta current.

Mixer

V<sub>CC1</sub>=V<sub>CC2</sub>=Enable=Mix Gain=2.75V; LNA Gain=0.0V

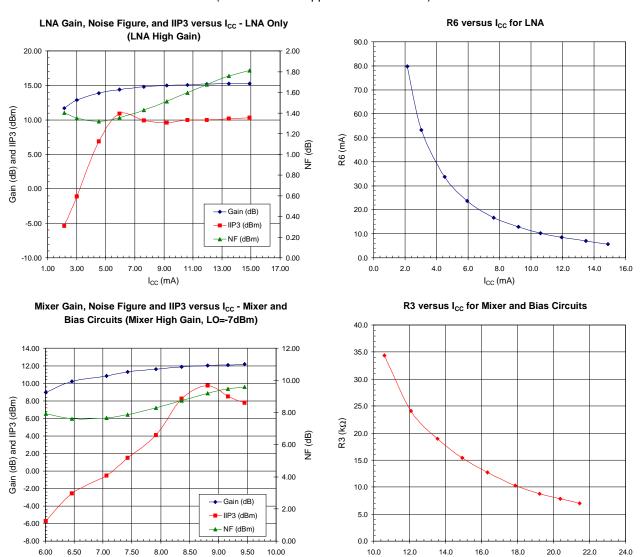
To measure I<sub>CC</sub> Mixer (LNA should be in bypass mode and LO signal should be present):

Total mixer current=I<sub>CC1</sub>

V<sub>CC2</sub> only affects LO current buffer and R6 doesn't affect the mixer current.

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W-CDMA (See W-CDMA Application Schematic)



**Instructions** (Board loss, taking into consideration description in the W-CDMA schematic) LNA

 $I_{CC}$  LNA current=total current ( $V_{CC}$ =LNA Gain=2.75)-total current ( $V_{CC}$ =2.75; LNA Gain=0) To measure  $I_{CC}$  LNA only:

LNA Gain was switched between 0V and 2.75V, and record the delta current.

 $I_{CC}$  (mA)

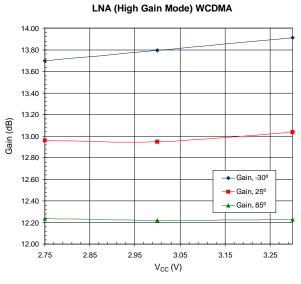
#### Mixer

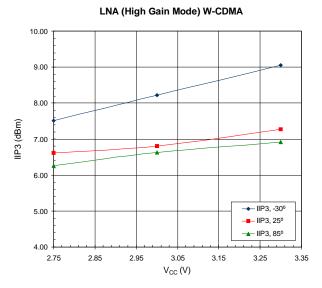
 $I_{CC}$  Mix and bias current=total current ( $V_{CC}$ ;=EN= $V_{CC2}$ =Mix Gain=2.75; LNA Gain=0)-total current ( $V_{CC}$ ;=EN=2.75; Mix Gain=LNA Gain= $V_{CC2}$ =0

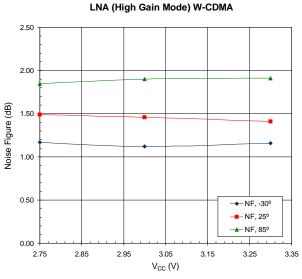
 $I_{CC}$  (mA)

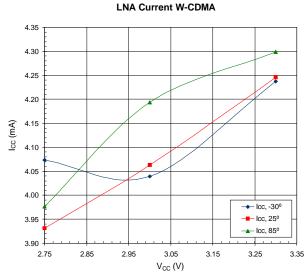
LO signal should be present.  $V_{\text{CC2}}$  only affects LO current buffer and R6 doesn't affect the mixer current.

By using a R6=39k $\Omega$  and R3=24k $\Omega$ , the following results were obtained. RF=2140MHz, LO=2330MHz, IF=190MHz.









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8.0

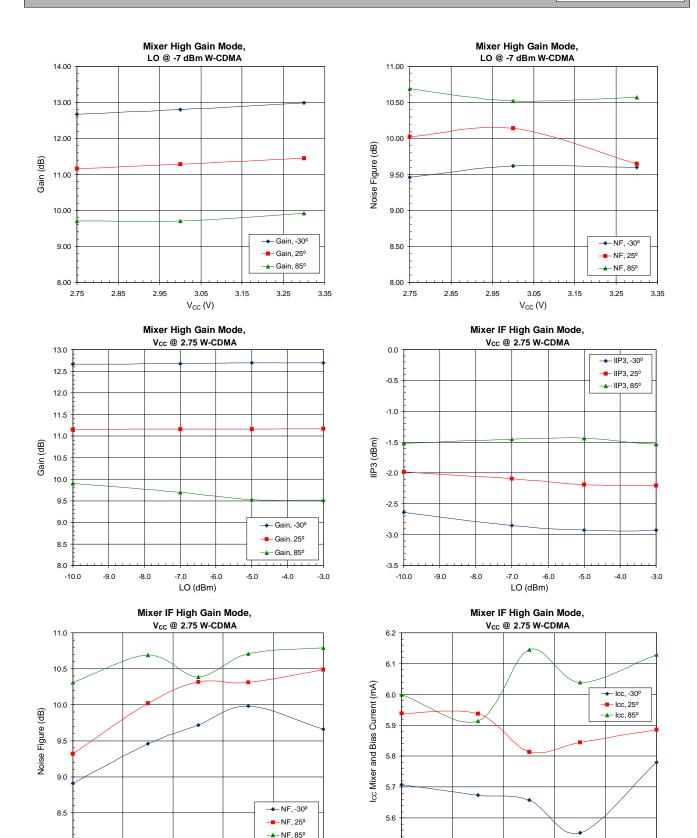
-10.0

-8.0

LO (dBm)

0.0

-2.0



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0.0

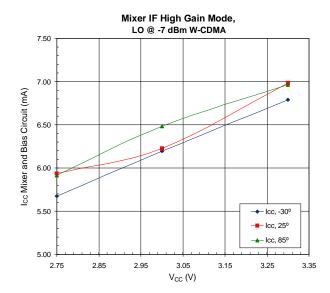
-2.0

5.5

-10.0

-8.0

LO (dBm)



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