## AN6227FHN

## Single chip, transmission and reception IC for PDC

### ■ Overview

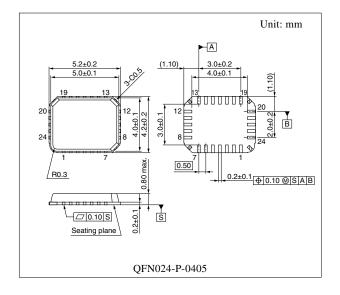
The AN6227FHN is a transmission and reception IC incorporating reception sleep function for a 1.5 GHz cellular telephone.

#### ■ Features

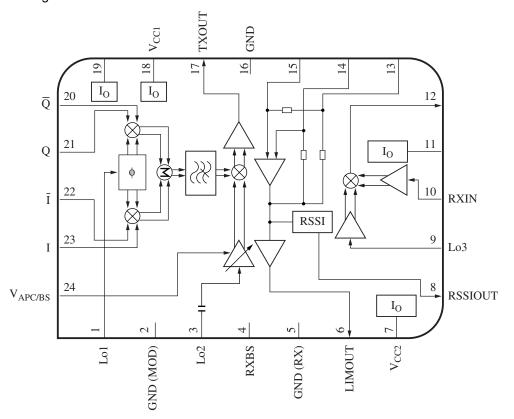
- Reception sleep function built-in
- Ultra mini-type 4 mm × 5 mm leadless package
- Current consumption: At reception: 25 mA
   At transmission: 3.2 mA

### ■ Applications

• Cellular telephone (1.5 GHz PDC)



### ■ Block Diagram



### ■ Pin Descriptions

Pin No.	Symbol	Description	Pin No.	Symbol	Description
1	TXLO1	TX local 1 input	13	LMDEC1	Limiter decouple 1
2	GNDMOD	TX modulator GND	14	LMDEC2	Limiter decouple 2
3	TXLO2	TX local 2	15	LMIN	Limiter input
4	RXBS	RXBS	16	GNDOUT	TX output GND
5	GNDRX	RX GND	17	TXOUT	TX output
6	LMOUT	Limiter output	18	VCCOUT	TX output V <sub>CC</sub>
7	VCCLIM	V <sub>CC</sub> limiter	19	VCCMOD	TX modulator V <sub>CC</sub>
8	RSOUT	RSSI output	20	Q-IN	Q input
9	RXLOIN	RX local input	21	Q-IN	Q input
10	RXMXIN	RX mixer input	22	Ī-IN	Ī input
11	VCCMIX	Mixer V <sub>CC</sub>	23	I-IN	I input
12	MXOUT	Mixer output	24	APC/BS	APC/BS

### ■ Absolute Maximum Ratings

Parameter	Symbol	Rating	Unit
Supply voltage	V <sub>CC</sub>	4.2	V
Supply current	I <sub>CC</sub>	60	mA
Power dissipation *2	$P_{\mathrm{D}}$	125	mW
Operating ambient temperature *1	T <sub>opr</sub>	-30 to +80	°C
Storage temperature *1	T <sub>stg</sub>	-55 to +125	°C

Note) \*1: Except for the operating ambient temperature and storage temperature, all ratings are for  $T_a = 25^{\circ}C$ .

### ■ Recommended Operating Range

Parameter	Symbol	Range	Unit
Supply voltage	V <sub>CC</sub>	2.6 to 4.0	V

### $\blacksquare$ Electrical Characteristics at $T_a=25^{\circ}C$

Parameter	Symbol	Conditions	Min	Тур	Max	Unit
Current consumption	I <sub>CCTX</sub>	Lo1 = 178 MHz, −25 dBm	_	25	33	mA
(transimisson) *1		Lo2 = 1619  MHz, -18  dBm				
		$V_{APC} = 2.3 \text{ V}$				
Sleep current *1	I <sub>SLTX</sub>	No signal, $V_{APC/BS} \le 0.3 \text{ V}$	_	0	10	μΑ

<sup>\*2:</sup>  $P_D$  is the value at  $T_a = 80^{\circ}C$  without a heatsink. Use this device within the range of allowable power dissipation referring to "Technical Data"  $P_D - T_a$  curves of QFN024-P-0405".

### ■ Electrical Characteristics at T<sub>a</sub> = 25°C (continued)

Parameter	Symbol	Conditions	Min	Тур	Max	Unit
Output level 1 *1	P <sub>O1</sub>	Lo1 = 178 MHz, -25 dBm	-16	-13	_	dBm
		Lo2 = 1607  MHz, -18  dBm				
0 4 41 12*	D	$V_{APC} = 2.3 \text{ V}$	16	12		ID
Output level 2 *1	P <sub>O2</sub>	Lo1 = 178 MHz, -25 dBm Lo2 = 1631 MHz, -18 dBm	-16	-13	_	dBm
		$V_{APC} = 2.3 \text{ V}$				
Minimum output level *1	P <sub>min</sub>	Lo1 = 178 MHz, -25 dBm	T —	-50	-40	dBm
		Lo2 = 1619  MHz, -18  dBm				
		$V_{APC} = 1.0 \text{ V}$				
Current consumption (reception) *2	I <sub>CCRX</sub>	No signal	_	3.2	4.5	mA
Reception sleep current *2	I <sub>RXSLP</sub>	No signal, RXBS ≤ 0.3 V	_	_	10	μA
Mixer conversion gain *2	$G_{MX}$	$V_{MI} = 60 \text{ dB}\mu$ , SW1 = b (refer to	20	23	26	dB
		"■ Application Circuit Example"),				
		Excludes the filter loss of -7 dB				
Mixer maximum output	V <sub>MX</sub>	$V_{MI} = 105 \text{ dB}\mu$ , SW1 = b (refer to	100	106	_	dΒμ
amplitude *2		"■ Application Circuit Example"),				
		Excludes the filter loss of -7 dB				
Limiter voltage gain *2	$G_{LM}$	$V_{LI} = 15 \text{ dB}\mu$	80	85	90	dB
Limiter maximum output amplitude *2	$V_{LM}$	$V_{LI} = 80 \text{ dB}\mu$ , 450 kHz component	0.90	1.25	1.60	V[p-p]
RSSI output voltage 1 *2	$V_{S(1)}$	$V_{LI} = 0 dB\mu$	0	0.23	0.6	V
RSSI output voltage 2 *2	V <sub>S(2)</sub>	$V_{LI} = 115 \text{ dB}\mu$	2.31	2.6	2.91	V
RSSI reference output slope *3	D <sub>S</sub>	$V_S (V_{IS}) = V_{S(I)} + 0.12 \text{ V}$ $D_S = V_S (V_{IS} + 75 \text{ dB}\mu) - V(V_{IS})$	1.39	1.8	2.19	V
RSSI output slope variation 1 *3	$\Delta D_{S(1)}$	$\Delta D_{S(1)} = 5 \{V_S (V_{IS} + 15 dB\mu) - V_S (V_{IS}) \} /D_S$	0.75	1	1.25	_
RSSI output slope variation 2 *3	$\Delta D_{S(2)}$	$\Delta D_{S(2)} = 5 \{V_S (V_{IS} + 30 \text{ dB}\mu) - V_S (V_{IS} + 15 \text{ dB}\mu)\} / D_S$	0.75	1	1.25	_
RSSI output slope variation 3 *3	$\Delta D_{S(3)}$	$\Delta D_{S(3)} = 5 \{V_S (V_{IS} + 45 dB\mu) - V_S (V_{IS} + 30 dB\mu)\} / D_S$	0.75	1	1.25	_
RSSI output slope variation 4 *3	$\Delta D_{S(4)}$	$\Delta D_{S(4)} = 5 \{V_S (V_{IS} + 60 \text{ dB}\mu) - V_S (V_{IS} + 45 \text{ dB}\mu)\} / D_S$	0.75	1	1.25	_
RSSI output slope variation 5 *3	$\Delta D_{S(5)}$	$\Delta D_{S(5)} = 5 \{V_S (V_{IS} + 75 dB\mu) - V_S (V_{IS} + 60 dB\mu)\} /D_S$	0.75	1	1.25	_

Note) \*1: V<sub>CC1</sub> = 3.0 V, IQ signal amplitude: 0.18 V[p-p] (both phases), DC bias: 1.6 V, (π/4 QPSK-modulated [0000] continuous wave input.

Output frequency of  $P_{O1}$ : 1429.0025 MHz, output frequency of  $P_{O2}$ : 1453.0025 Hz, output frequency of  $P_{min}$ : 1441.0025 MHz. Output level is measured with a spectrum analyzer.

Setting of a spectrum analyzer: SPAN = 20 kHz, RBW = 300 Hz, VBW = 30 Hz, ST = 5 s

(When inputting  $\pi/4$  QPSK-modulated [0000] continuous wave as IQ signal, the frequency for  $P_{O1}$ ,  $P_{O2}$  and  $P_{min}$  becomes Lo frequency plus IQ signal frequency, which leads to the above value.)

Lo input level is a setting value of signal source (output impedance 50 Ω) described in the "■ Application Circuit Example".

### ■ Electrical Characteristics at T<sub>a</sub> = 25°C (continued)

Note) (continued)

\*2: Unless otherwise specified:  $V_{CC2} = 3.0 \text{ V}$ , RXBS = 2.5 V to 3.0 V, SW1 = a (Refer to "  $\blacksquare$  Application Circuit Example").  $V_{LO3} = 90 \text{ dB}\mu$ : f = 129.55 MHz,  $V_{MI}$ : f = 130 MHz,  $V_{LI}$ : f = 450 kHz

(Input level of pin 15 is excluded the loss of the matching circuit and filter.)

 $V_{MX}$  and  $V_{LM}$  are measured in high impedance.

Lo input level is a setting value of signal source (output impedance 50 Ω) described in the "■ Application Circuit Example".

\*3:  $V_{IS}$  is the input level  $V_{L1}$  at which the RSSI output voltage becomes  $V_{S(1)} + 0.12 \text{ V}$ .

### • Design reference data

Unless otherwise specified,  $V_{CC1} = 3.0 \text{ V}$ .

Lo input level is a setting value of signal source (output impedance  $50 \Omega$ ) described in the "Application Circuit Example". Note) The characteristics listed below are theoretical values based on the IC design and are not guaranteed.

Parameter	Symbol	Conditions	Min	Тур	Max	Unit
Carrier leak suppression *1	CL	Lo1 = 178 MHz, -25 dBm	_	-35	-25	dBc
(fLo2-fLo1)		Lo2 = 1619  MHz, -18 dBm				
		$V_{APC} = 2.3 \text{ V}$				
Image leak suppression *1	IL	Lo1 = 178  MHz, -25  dBm	_	-35	-30	dBc
		Lo2 = 1619  MHz, -18  dBm				
		$V_{APC} = 2.3 \text{ V}$				
Proximity spurious suppression *1	DU	Lo1 = 178 MHz, -25 dBm	_	-70	-65	dBc
		Lo2 = 1619  MHz, -18  dBm				
B 1 1 1 2 2 2 2 1	D.D.	$V_{APC} = 2.3 \text{ V}$		40	20	
Base band distortion suppression *1	BD	Lo1 = 178 MHz, -25 dBm	_	-40	-30	dBc
		Lo2 = 1619  MHz, -18  dBm $V_{APC} = 2.3 \text{ V}$				
A.P. (1 11 1	DI 1			15	20	ID
Adjacent channel leak power suppression (30 kHz detuning) *2	BL1	Lo1 = 178 MHz, -25 dBm Lo2 = 1619 MHz, -18 dBm	_	<del>-45</del>	-38	dBc
suppression (50 KHZ detuining)		$V_{APC} = 2.3 \text{ V}$				
Adjacent channel leak power	BL2	Lo1 = 178 MHz, -25 dBm		-70	-60	dBc
suppression (50 kHz detuning) *2	BL2	Lo2 = 1619  MHz, -18  dBm	_	-70	-00	ubc
suppression (50 kHz detailing)		$V_{APC} = 2.3 \text{ V}$				
Adjacent channel leak power	BL3	Lo1 = 178 MHz, -25 dBm			-65	dBc
suppression (100 kHz detuning) *2		Lo2 = 1619  MHz, -18  dBm				
		$V_{APC} = 2.3 \text{ V}$				
APC variable width *1	L <sub>APC</sub>	Lo1 = 178 MHz, -25 dBm	30	37	45	dB
		Lo2 = 1619  MHz, -18  dBm				
		$V_{APC} = 1.0 \text{ V to } 2.3 \text{ V}$				
APC output level control	S <sub>APC</sub>	Lo1 = 178 MHz, -25 dBm	37	46	55	dB/V
sensitivity *1		Lo2 = 1619  MHz, -18  dBm				
		$V_{APC} = 1.0 \text{ V}/1.6 \text{ V}$				
In-band output level deviation *1	ΔΡ	Lo1 = 178 MHz, -25 dBm	-1.5	_	+1.5	dB
		Lo2 = 1607  MHz to  1631  MHz,				
		$-18 \text{ dBm}, V_{APC} = 2.3 \text{ V}$				

### ■ Electrical Characteristics at $T_a = 25$ °C (continued)

#### • Design reference data (continued)

Unless otherwise specified,  $V_{CC1} = 3.0 \text{ V}$ .

Lo input level is a setting value of signal source (output impedance  $50 \Omega$ ) described in the " $\blacksquare$  Application Circuit Example".

Note) The characteristics listed below are theoretical values based on the IC design and are not guaranteed.

Parameter	Symbol	Conditions	Min	Тур	Max	Unit
Modulation precision *3	EVM	Lo1 = 178 MHz, -25 dBm	_	2.0	3.5	%[rms]
		Lo2 = 1619  MHz, -18  dBm				
		$V_{APC} = 2.3 \text{ V}$				

Note) \*1: IQ signal amplitude: 0.18 V[p-p] (both phases), DC bias: 1.6 V, π/4 QPSK-modulated [0000] continuous wave input. Measure the suppression amount for output with a spectrum analyzer.

Setting of a spectrum analyzer: SPAN = 20 kHz, RBW = 300 Hz, VBW = 30 Hz, ST = 5 s

- \*2: IQ signal amplitude: 0.18 V[p-p] (both phases), DC bias: 1.6 V,  $\pi/4$  QPSK-modulated [PN9] continuous wave input. To be measured by a spectrum analyzer. (By using a leak power measurement function for an adjacent channel.) Setting of a spectrum analyzer: SPAN = 250 kHz, RBW = 1 kHz, VBW = 1 kHz, ST = 2 s
- \*3: IQ signal amplitude: 0.18 V[p-p] (both phases), DC bias: 1.6 V, π/4 QPSK-modulated [PN9] continuous wave input. The output level be measured by a spectrum analyzer. (By using a modulation precision measurement function.)

### ■ Terminal Equivalent Circuits

Pin No.	Equivalent circuit	[	Description	I/O
1	$ \begin{array}{c c}  & 19 \\ \hline  & 10 \text{ k}\Omega \\ \hline  & 2 \text{ pF 2 pF} & 450 \Omega \end{array} $	TXLO1: Input pin of qu	adrature modulator.	I
2	2 pF 2 pF 450 Ω 2 pF 2 pF 450 Ω 2 pF 2 pF 2 pF 450 Ω		ase shifter and modulator.	_
3	$ \begin{array}{c c} \hline 1 & k\Omega \\ \hline 1 & k\Omega \end{array} $	TXLO2: Local input pir	n for up mixer.	I
4	$\begin{array}{c c} \hline Regulator \\ \hline 200 \\ k\Omega \\ \hline \end{array}$	RXBS: On/off control RXBS (V) 0 to 0.3 2.5 to 3	pin for reception block.  Reception block  Off  On	I

### ■ Terminal Equivalent Circuits (continued)

Pin No.	Equivalent circuit	Description	I/O
5	7	GNDRX: GND pin of reception system. Make impedance low by widening the GND pattern.	_
6	180 μΑ	LMOUT: Output pin of limiter amplifier.	О
7		VCCLIM: $V_{CC}$ pin for IF limiter amplifier RSSI.	_
8	7 23 kΩ 8 5 =	RSOUT: RSSI output pin. DC potential corresponding to input signal level of limiter amplifier is outputted.	O
9	$\begin{array}{c c} \hline 11 \\ \hline 11.2 \text{ k}\Omega \\ \hline 5 \text{ k}\Omega \\ \hline 5 \text{ k}\Omega \end{array}$	RXLOIN: Local input pin for reception down mixer.	Ī
10		RXMXIN: Input pin to 1st. IF amplifier. Input impedance is $2 \text{ k}\Omega$ .	I
11	$\begin{array}{c c} 1 & k\Omega \\ 1 & k\Omega \\ \end{array}$	VCCMIX: $V_{CC}$ pin for reception down mixer.	_
12	360 µA 8 5 =	MXOUT: Reception down-mixer output pin.	O

### ■ Terminal Equivalent Circuits (continued)

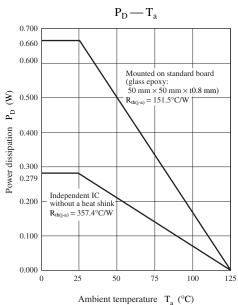
Pin No.	Equivalent circuit	Description	I/O
13, 14	(3) (14) (8) 2 kΩ 100 kΩ (10) = (10) kΩ	Pin 13: LMDEC1; Pin 14: LMDEC2: De-coupling pin for feedback of limiter amplifier. Connect an external capacitor to GND.	_
15	8.5 kΩ 102 kΩ 5	LMIN: Limiter amplifier input pin. Input impedance is $2 \text{ k}\Omega$ .	I
16	18	GNDOUT: GND pin for transmission up-mixer and RF output amplifier.	_
17	17	TXOUT: RF output pin from output amplifier circuit.	O
18	16	VCCOUT: $V_{CC}$ pin for transmission up-mixer and RF output amplifier.	_
19		$\label{eq:VCCMOD:} V_{CC} \mbox{ pin for phase shifter and quadrature } \\ \mbox{modulator.}$	_
20	19 (600 Ω 200 Ω 200 Ω 600 Ω 60	$\begin{tabular}{ c c c c c c c c c c c c c c c c c c c$	I
21	21 + 600 Ω 200 Ω 200 Ω 200 Ω 200 Ω 200 Ω 3 pF 3 pF 3 pF 5 5	$\begin{array}{c} \text{Q-IN:} \\ \text{Q signal input pin.} \\ \text{Relation between DC bias and amplitude} \\ \text{is as follows:} \\ \hline \\ \text{DC bias (V)} \\ \hline \\ \text{Amplitude (V[p-p])} \\ \hline \\ \text{1.6} \\ \hline \\ \text{Input impedance is } 100 \text{ k}\Omega \text{ or more.} \\ \hline \end{array}$	I

### ■ Terminal Equivalent Circuits (continued)

Pin No.	Equivalent circuit	Description	I/O
22	19	Ī-IN: Ī signal input pin. Relation between DC bias and amplitude is as follows:	I
		$\begin{tabular}{ c c c c c c c c c c c c c c c c c c c$	
23	22 + Ω 200 Ω 600 Ω 233 pF 3 pF 5 5	I-IN: I signal input pin. Relation between DC bias and amplitude is as follows:  DC bias (V)   Amplitude (V[p-p])   1.6   0.18	I
		Input impedance is $100 \text{ k}\Omega$ or more.	
24	Regulator $ \begin{array}{c} 200 \text{ k}\Omega \\ 10 \text{ k}\Omega \end{array} $ (APC control)	APC/BC: Pin for use both as battery saving of transmission block and as power control of transmitting RF output.  Control with the following conditions: $V_{APC}(V)$ Mode  0 to 0.3  Off  1.0 to $V_{CC}$ On (APC control)	I
	(APC control)	Input impedance is 5 k $\Omega$ or more.	

### ■ Technical Data

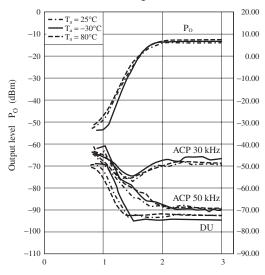
### 1. $P_D - T_a$ curves of QFN024-P-0405



### ■ Technical Data (continued)

#### 2. Main characteristics

#### APC control voltage characteristics



APC control voltage (V)

 $V_{CC} = 3.0 \text{ V}, T_a = -30^{\circ}\text{C}, 25^{\circ}\text{C}, 80^{\circ}\text{C},$ 

 $BS = V_{APC} = V_{AR}$ 

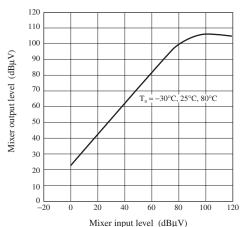
Lo1: 178 MHz, -25 dBm

Lo2: 1619 MHz, -18 dBm

I, Q : 0.18 V[p-p] (both phases) 1.6 V $_{DC}$  ,  $\pi/4$ , [0000] or using PN9 stages continuous wave.

Adjacent channel leak power suppression amount: ACP 30 kHz, ACP 50 kHz (dBc) Proximity spurious suppression amount: DU (dBc)

#### Mixer characteristic

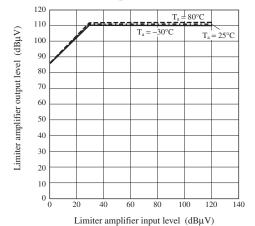


 $V_{CC} = 3.0 \text{ V}, T_a = -30^{\circ}\text{C}, 25^{\circ}\text{C}, 80^{\circ}\text{C}$ 

Mixer in: 130 MHz Mixer out: 450 kHz

Lo3 in: 129.55 MHz, 90 dBµV

#### Limiter amplifier characteristics

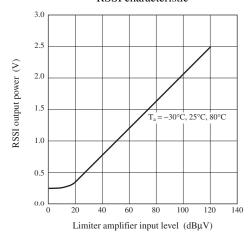


 $V_{CC} = 3.0 \text{ V}, T_a = -30^{\circ}\text{C}, 25^{\circ}\text{C}, 80^{\circ}\text{C},$ 

BS = 2.5 V

Limiter in: 450 MHz, Limiter out: 450 kHz

### RSSI characteristic

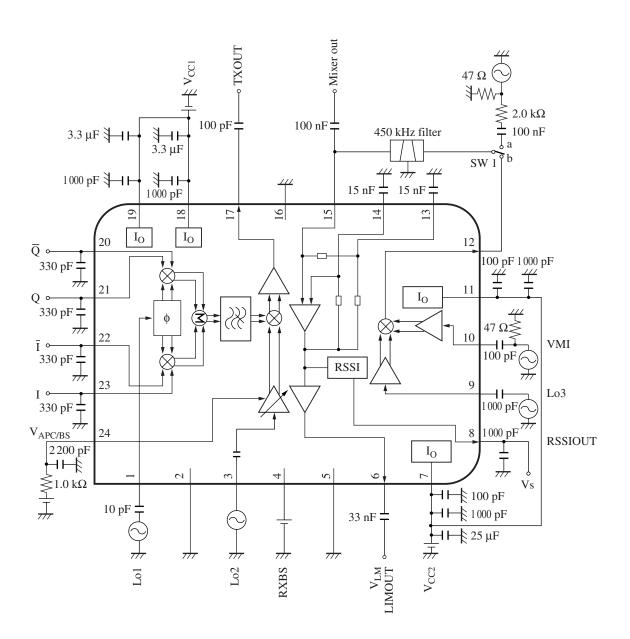


 $V_{CC} = 3.0 \text{ V}, T_a = -30^{\circ}\text{C}, 25^{\circ}\text{C}, 80^{\circ}\text{C},$ 

BS = 2.5 V

Limiter in: 450 MHz, Limiter out: 450 kHz

### ■ Application Circuit Example



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